

RESEARCH ARTICLE

Numerical Exposure Assessment Method for Low Frequency Range and Application to Wireless Power Transfer

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Abstract

In this paper, a numerical exposure assessment method is presented for a quasi-static analysis by the use of finite-difference time-domain (FDTD) algorithm. The proposed method is composed of scattered field FDTD method and quasi-static approximation for analyzing of the low frequency band electromagnetic problems. The proposed method provides an effective tool to compute induced electric fields in an anatomically realistic human voxel model exposed to an arbitrary non-uniform field source in the low frequency ranges. The method is verified, and excellent agreement with theoretical solutions is found for a dielectric sphere model exposed to a magnetic dipole source. The assessment method serves a practical example of the electric fields, current densities, and specific absorption rates induced in a human head and body in close proximity to a 150-kHz wireless power transfer system for cell phone charging. The results are compared to the limits recommended by the International Commission on Non-Ionizing Radiation Protection (ICNIRP) and the IEEE standard guidelines.

Introduction

The concept of wireless power transfer (WPT) technique has been firstly proposed by Nikolas Tesla a century ago. Since a WPT technique using high resonance has been reported by MIT (Massachusetts Institute of Technology) research team in 2007 [1], many researchers have been studying on the WPT technique. Such studies were summarized in [2], and historical review of this technique was outlined as WPT has been researched applications for medical implants, induction heaters, inductive power transfer systems, and wireless charging systems for portable equipment [3–11]. Recently, WPT technique for cellular phone charging has been commercialized by the Qi standard of the Wireless Power Consortium [12]. The radiation level of electric, magnetic, and electromagnetic fields (EMFs) generated from the WPT system is extremely strong compared to that of EMFs from wireless communications. According to the specifications of the Wireless Power Consortium and AirFuel Alliance [13], charging

powers are 5W to 15W for mobile devices such as cell phones and tablets, up to 100W for consumer electronics, and up to 2kW for kitchen utensils. Moreover, for the WPT applications of electric vehicles, charging powers can range from 3.3 kW to 22kW for light duty vehicles and reach 100kW for heavy duty vehicles. With continued increases in transmitted power level and transmission distance for WPT through the air, we should carefully consider electromagnetic compatibility (EMC) and the adverse health effects of EMFs exposure in close proximity to a WPT system [14].

To consider the safety issues related to human exposure to EMFs generated from the WPT system, we should evaluate internal electric fields in a human body and confirm compliance with international safety guidelines [15–18]. Due to the difficulty of measurement, numerical methods are often used for exposure assessment with anatomical human body models [19–22]. For computation of the bioelectric field, the full-wave finite-difference time-domain (FDTD) method [23] is commonly used to evaluate the internal electric field of various tissues in a human body at radio frequencies. For mid-range wireless power research, the quality factor for both transmitting and receiving resonant circuits should be large [1], and operating a WPT system at a higher resonant frequency is one of the approaches utilized to have a high quality factor. However, the operating frequencies of approximately few and/or hundreds of kHz band are discussed due to the energy efficiency and costs of overall system. Moreover, the lower operating frequency has the benefit of compliance with international guidelines because the reference levels for exposure to time-varying EMFs decrease with frequency to around 10 MHz. For operating frequency, the Wireless Power Consortium adopts a 110 kHz to 205 kHz range, and AirFuel Alliance adopts around 115 kHz, 227 kHz, and 6.78 MHz. Operating frequency bands of 20 kHz, 60 kHz, and 85 kHz are indicated for electric vehicle charging. Therefore, it is difficult to apply the conventional FDTD method to dosimetry for WPT systems in this frequency range due to the increased number of time steps. For numerical dosimetry for WPT systems, a numerical technique for a low frequency band is substantial. Various exposure scenarios should be evaluated and the fast algorithm will be helpful to process big data [24–26].

The impedance [27] and scalar-potential finite-difference (SPFD) methods [28] based on the quasi-static (QS) approximation have been used at low frequencies (frequencies up to a few megahertz). The previous works [29–32] on numerical dosimetry for approximately 10 MHz operating WPT systems have used full-wave analysis (FDTD method) and QS analysis (impedance and SPFD methods considering only incident magnetic fields) because both are available in this frequency range. As previous descriptions have indicated, a low-frequency analysis should be conducted for dosimetry for WPT systems because the operating frequency for nearly all WPT systems is in the range of a few tens of kilohertz to few hundreds of kilohertz. The QS-FDTD method has been used to compute induced currents in human models for the WPT frequency range (100 kHz to 10 MHz) [33]. However, the models were exposed to only uniform magnetic fields, in which two plane waves are superposed with different polarizations and propagation directions to cancel out electric fields, while non-uniform electric and magnetic fields are generated from a WPT system.

In this paper, a numerical exposure assessment method is proposed for low frequency range analysis. To verify our method, the internal electric field in a sphere model as a simplified human phantom model exposed to magnetic dipole is obtained by the proposed method and compared with theoretical results. Moreover, a WPT system for cell phone charging operating at 150 kHz frequency is designed, and dosimetry for the system is conducted with a human body model. Finally, we discuss the compliance of the WPT system with international safety guidelines.

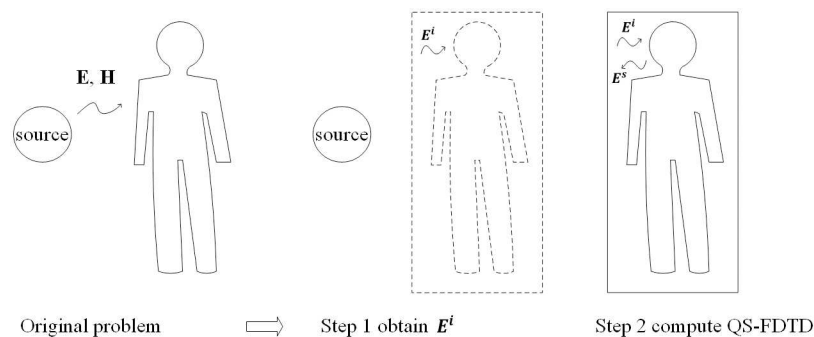


Fig 1. Illustration of two-step process to compute a bioelectric field problem.

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Numerical Dosimetry Method

Theory

The proposed numerical dosimetry method is conducted by a two-step process, similar to the previous work [31], instead of using full-wave analysis for the original problem all at once, as shown in Fig 1. The first step is to obtain the electric field generated from a source, which will be generated from a source, such as a WPT system, power line, or induction heater in the absence of a human body. The electric field can be obtained by any theoretical or numerical methods and should be located in the space occupied by a human body. In the second step, the induced electric field in a human body is calculated with a scattered-field FDTD method [23] by regarding the electric field obtained in the previous step as the incident electric field to the human body. It is noted that total electric and magnetic fields can be obtained with only the incident electric field, without the incident magnetic field in the scattered-field FDTD method. At this time, however, the QS-FDTD method is employed [34] instead of the conventional FDTD method for low-frequency analysis.

Using the QS-FDTD method, the number of time steps can be considerably decreased due to rapid convergence within a much shorter time than one full period in the quasi-static system, whereas the conventional FDTD method has to simulate several periods to reach the steady state. Under the quasi-static condition, the external fields outside a human body all have the same phase as the incident field. However, the internal fields inside a human body are proportional to the time derivative of the incident field. If a ramp function is used as an incident field, the external fields change linearly, whereas the internal fields remain constant. To avoid high-frequency contamination of the incident field, the following modified ramp function is generally used:

$$E_{inc} = \begin{cases} 0, & -\infty < t \leq t_0 \\ \cosh(t - t_0), & t_0 < t \leq \tau \\ \alpha \times (t - \tau), & t > \tau \end{cases} \quad (1)$$

In this modified ramp function, α is the desired slope, which is related to the peak amplitude and frequency of the sinusoid wave source. The constant t_0 is the starting time of the ramp excitation function, and the constant τ is generally set by hundred time steps. However, it is noted that the computation of the QS-FDTD method is simply implemented by using the same algorithm (code) as the FDTD method.

The unique feature of the proposed method, as compared to previous research [33–35], is that the incident field can not only be uniform fields but also arbitrary non-uniform fields by

using a two-step process. A key concept of this method is that the non-uniform electric fields of a system desired for exposure assessment in the first step will be used as the incident fields in the second step. To describe these non-uniform fields as the incident fields, we should consider every different magnitude of the non-uniform field. The flowchart of our approach can be expressed as in Fig 2.

Verification

The main goal of this work is to estimate internal electric fields in a human body exposed to a WPT system at a low operating. Thus, we chose a source of magnetic dipole having EMF distribution similar to that of a WPT system. The internal electric fields in a sphere model as a simplified human phantom exposed to the magnetic dipole are calculated by the proposed method and compared to Mie theory's calculation [36]. The relative and absolute errors are investigated to verify our method and to find out locations of maximum error.

Fig 3 shows the sketch of a sphere model in reference to a magnetic dipole. The distance between the center of the sphere model and the magnetic dipole is 400 mm. The radius of the sphere is 100 mm, and the magnetic dipole is in the z-polarization. The relative dielectric constant and conductivity of the sphere are 40 and 1 S/m, respectively. Through this paper, all simulations were conducted with a grid of 2 mm resolution. Convergence of this calculation was reached after 2000 time steps. Fig 4 shows the internal electric field strengths by the proposed method at cross sections in three directions, along with the corresponding theoretical solution for comparison. It is clearly seen that the field spatial distributions throughout the sphere are in good agreement with the theoretical solution.

Table 1 provides point-by-point relative and absolute errors for maximum and average field. Relative errors of maximum and average field are 1.72% and 0.01%, respectively. Relative errors at interior points are generally less than 1%. Fig 5 shows the relative error distributions at the cross sections. We can find the high relative errors concentrated at the surface of the sphere due to the staircase error. Relative errors range from 20% to 50% at the surface of the sphere. We can also find very large relative errors around the center related to zero field value in the theoretical solution. It is not substantial for human exposure assessment, however, because these field values are very small. The calculated values close to boundaries and zero values in the theoretical solution are overestimated, but by combining the qualitative and quantitative results, the proposed method accurately predicts the induced field distribution and field values in the interior of a lossy dielectric object such as a human body.

Human Exposure Assessment for a WPT System

In this section, dosimetry for a WPT system is conducted by using the proposed method. The size and operating frequency of the WPT system are presently designed by referring to the uniquely commercialized Qi standard. Exposure assessment for the system is computed with head and whole body of human voxel model. The results are discussed in regards to international safety guidelines.

WPT system

Fig 6 shows the WPT system designed for cell phone charging. A practical WPT system uses ferrite sheet for electromagnetic interference (EMI) shielding to prevent eddy currents on it. However, in order to consider the worst exposure case, our system is simply designed without the EMI shielding. According to Qi standard, the operating frequency band ranges from 110 kHz to 205 kHz. Thus we added a lumped capacitance to the coil to operate (resonate) at 150 kHz in the middle of the frequency range. The diameter of coil was selected to be 100 mm with

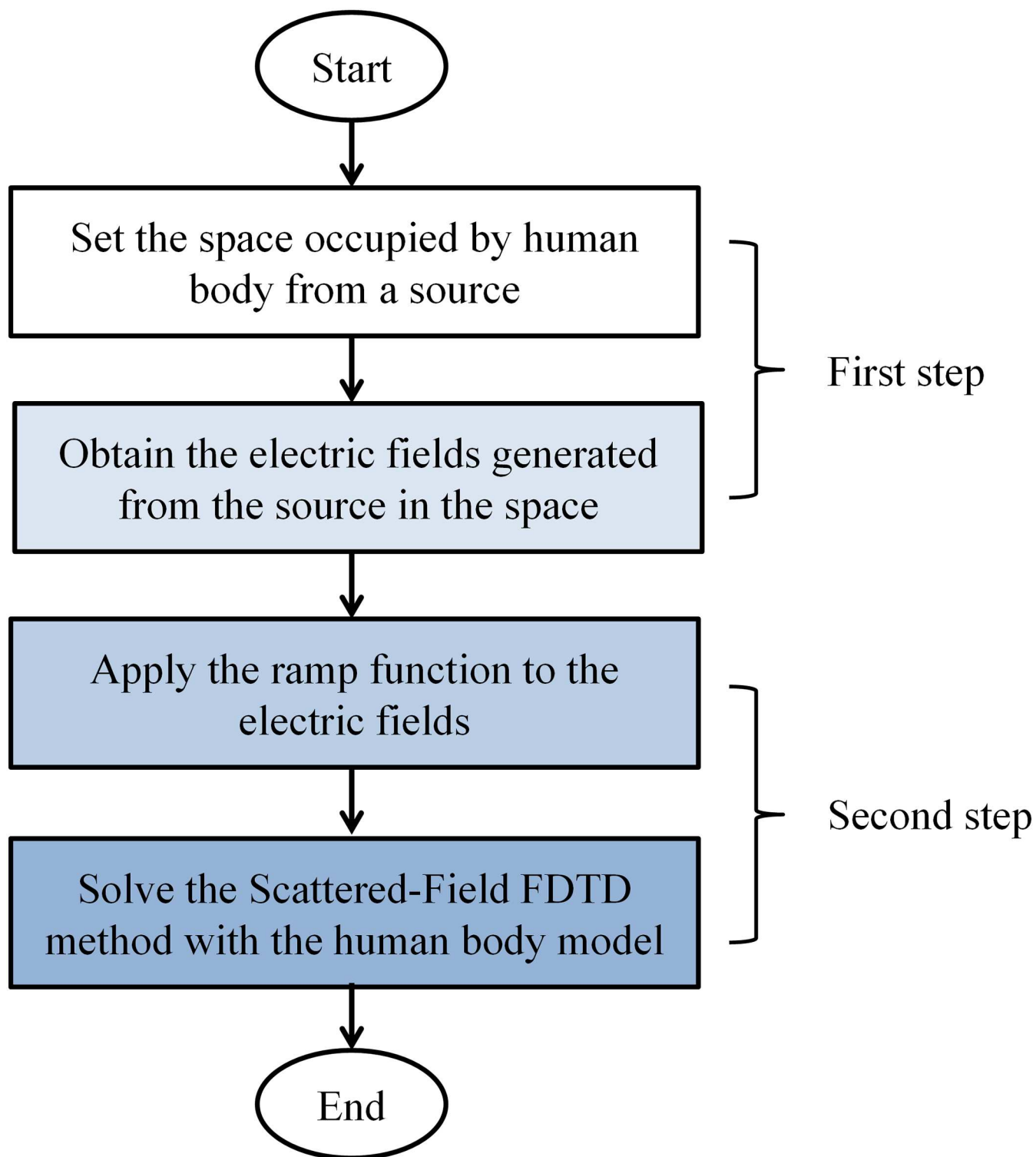


Fig 2. Flowchart of the proposed method.

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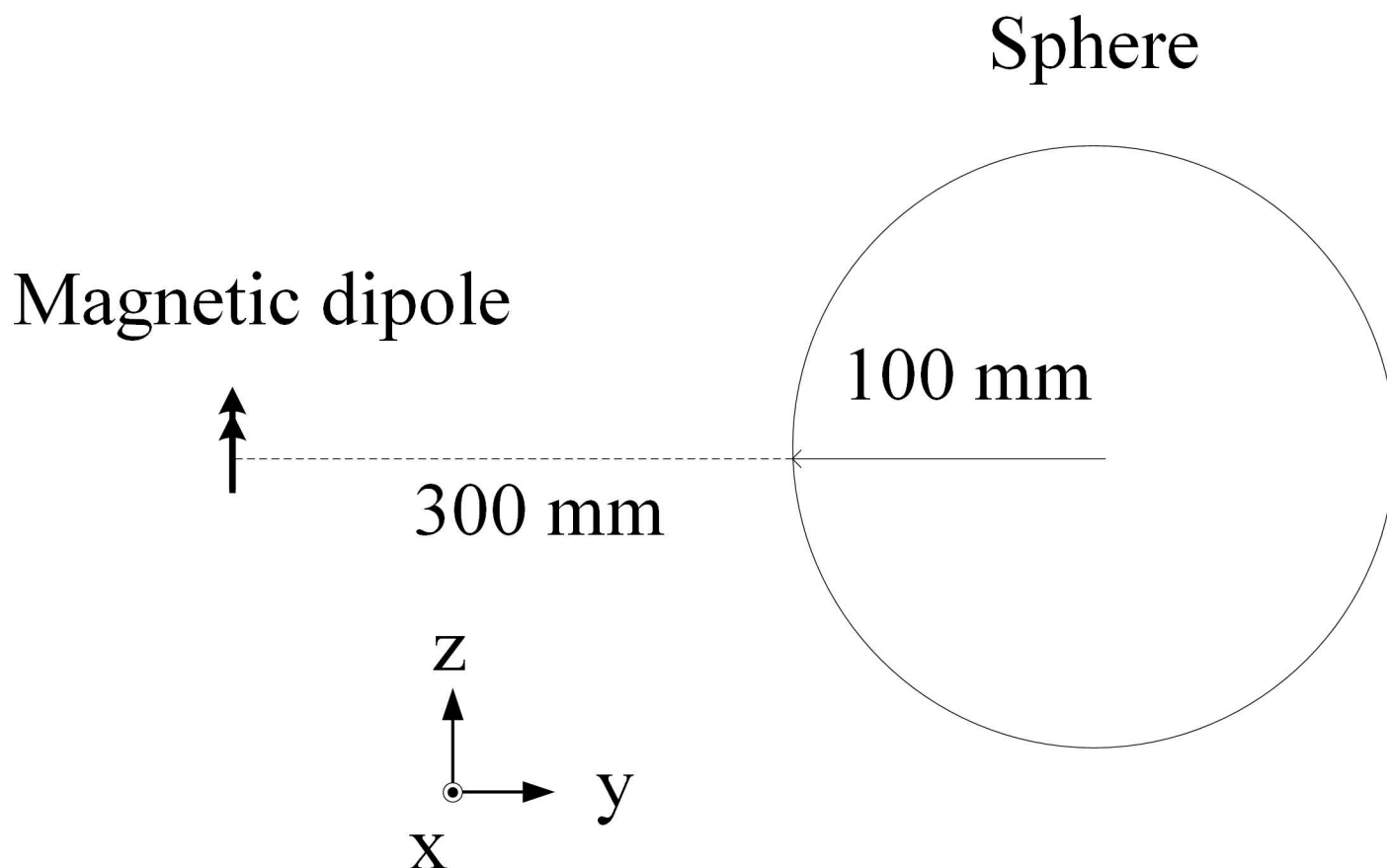


Fig 3. Magnetic dipole in reference to the lossy dielectric sphere.

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consideration to the size of a cell phone. Table 2 provides the parameters of the system. The transmitting coil and receiving coil are symmetrically designed, and the power transfer distance is set by 5 mm. Both ports are terminated with 50Ω . In this condition, the power transfer efficiency is about 90%.

International safety guidelines

In the International Commission on Non-Ionizing Radiation Protection (ICNIRP) and the IEEE standard guidelines up to 300 GHz, established adverse health effects depend on the frequency and intensity of the EMF. The electrostimulation effects dominate at low frequency, and the thermal effects dominate at high frequency. Thus, in the intermediate frequency region, we have to consider both effects. To protect against both effects, the ICNIRP and the IEEE standard recommend limits in the region from 100 kHz to 10 MHz and in the region from 100 kHz to 5 MHz, respectively. It is noted that in 2010, new ICNIRP guidelines (1 Hz to 100 kHz) were published to replace the low-frequency part of the 1998 ICNIRP guidelines (up to 300 GHz). The 99th percentile value of the internal electric field for a specific tissue (E_{99}), the induced current density (J)—which is used in ICNIRP guidelines—and the average value of the electric fields over a 5-mm-long straight-line segment oriented in any direction within the tissue (E_{5mm})—which is used in the IEEE standard—are the metrics for human exposure limits to protect against electrostimulation effects. The specific absorption rate (SAR) is used as the metric for human exposure limits to protect against thermal effects in both guidelines.

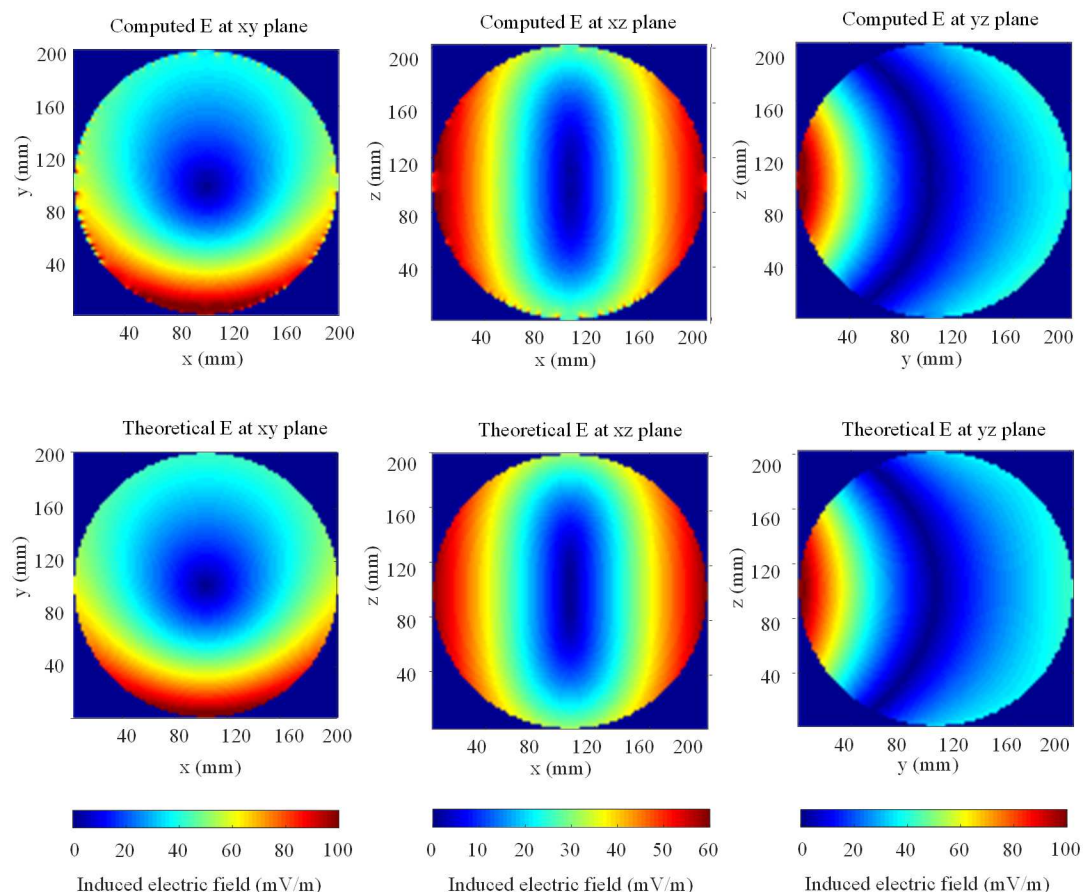


Fig 4. Internal electric field distribution in a sphere model exposed to a 1-MHz magnetic dipole for cross sections (computed vs theoretical). The upper and under figures show the computed and theoretical results, respectively.

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These metrics are called basic restrictions, which refer to restrictions on exposure to time-varying EMFs based directly on established health effects. Table 3 provides the basic restriction at 150 kHz of the WPT operating frequency.

Human model and exposure conditions

A whole-body voxel human model is used for dosimetry for a WPT system. The heterogeneous model of the human body TARO [20] is based on the accumulated magnetic resonant imaging

Table 1. Theoretical vs Computed electric field in a sphere model exposed to a 1-MHz magnetic dipole.

| | Maximum | Average |
|-------------------------------------|---------|---------|
| Theoretical electric field (mV/m) | 104.908 | 25.933 |
| Computed electric field (mV/m) | 106.709 | 25.935 |
| Relative error ^a (%) | 1.72 | 0.01 |
| Absolute error ^b (μmV/m) | 1.801 | 2.45e-3 |

^aDifference calculated by $\left| \frac{\text{Theory} - \text{Computation}}{\text{Theory}} \right| \times 100$.

^bError field calculated by $|\text{Theory} - \text{Computation}|$.

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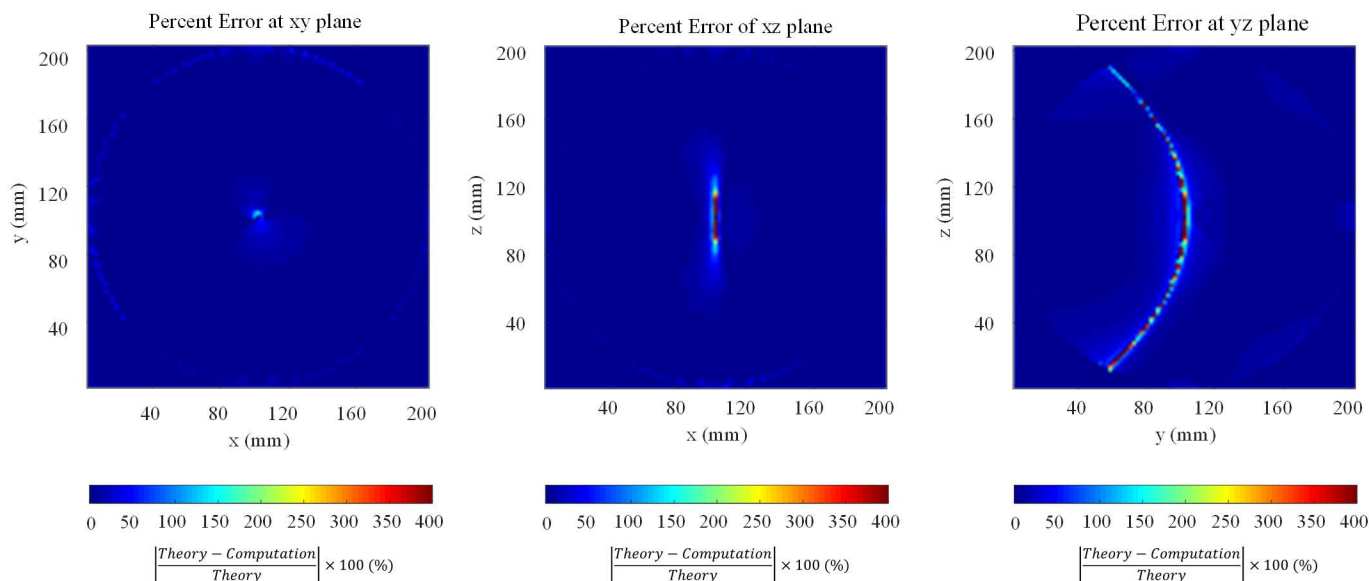


Fig 5. Relative error distribution at the cross sections.

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(MRI) data of adult Japanese volunteers. This anatomically realistic Japanese adult male model possesses 2-mm spatial resolution and 51 tissues and organs. In all computations, the x-axis of the coordinate system is from left to right, the y-axis is from back to front, and the z-axis is from foot to head as $320 \times 160 \times 866$ voxels. The various tissues and organs were assigned from Gabriel's Cole-Cole models [37].

Two exposure conditions are computed as illustrated in Fig 7. The head and whole body of TARO model are situated in front of the WPT system designed in this paper. The distance between the two centers of the WPT and TARO is 310 mm. Only the head model and the head of the whole-body model are located at the exact same position. All induced quantities were computed for an input power of 1 W.

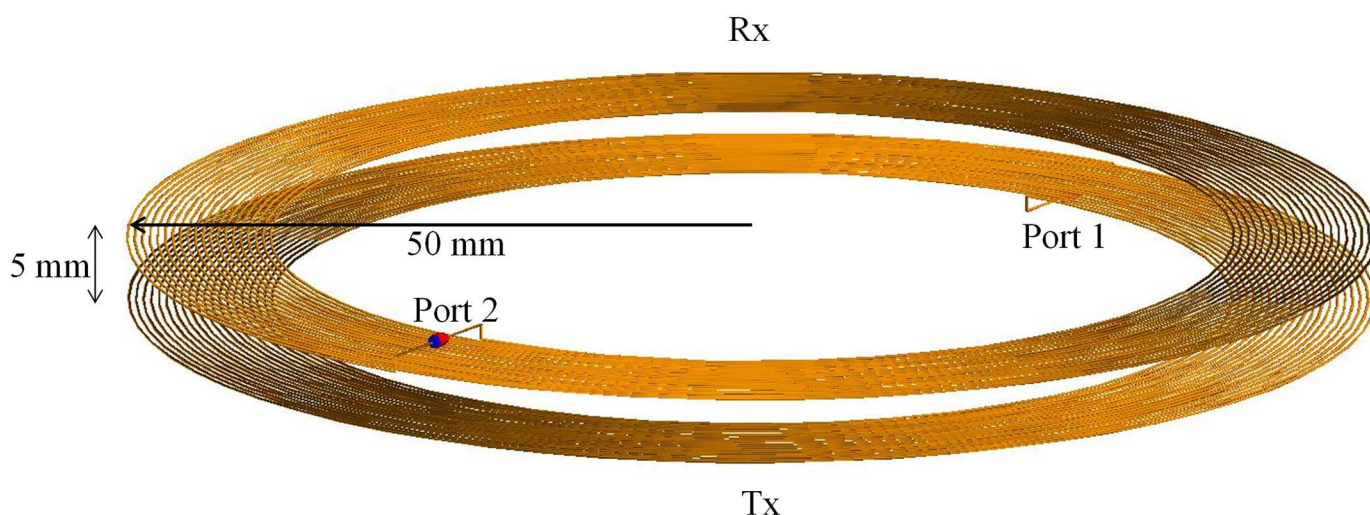


Fig 6. Configuration of transmitting and receiving coils for wireless power transfer.

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Table 2. Design parameters for a WPT system.

| | |
|------------------------|---------------|
| Operating frequency | 150 kHz |
| Coil inductance | 0.412 μ H |
| Resonance capacitance | 0.273 nF |
| Number of turns | 20 |
| Coil diameter | 100 mm |
| Distance between coils | 5 mm |

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Dosimetry results and discussion

The incident electric fields generated from the WPT system are computed by using the method of moment (MoM) [38]. The internal electric field, induced current density, and SAR are evaluated based on the proposed method. Steady state for the head and whole-body models were reached after 2000 and 3000 time steps, respectively. In the conventional FDTD method, the duration of one time step would be $\Delta t = \Delta x / \sqrt{3}c = 0.38$ ps with spatial discretization of $\Delta x = \Delta y = \Delta z = 0.2$ mm at 150 kHz from the Courant stability criterion. Generally, the conventional FDTD method requires a few periods of the source field. The required number of the time steps for four periods would be approximately 1.7×10^7 . In this frequency, the proposed method reduced the simulation time by nearly 5773-fold compared to the conventional FDTD method. However, the required memory for the proposed method is the same as that of the conventional FDTD method because they use the same algorithm.

Fig 8 shows the electric field, current density, and SAR distributions for the head and whole-body models. The two distributions for the head and the whole body are slightly different. Table 4 provides the induced quantities for the head and whole body. We calculated not only the metrics suggested by the international guidelines but also the induced quantities in the central nervous system (CNS)—which is important for the electrostimulation effects—and the localized SAR average of any 1 g cubical volume of tissue (SAR_{1g}). The results show that the induced quantities in the whole body are larger than those in the head only except for the SAR_{wm} due to the larger circular conductive loop in the whole body compared to that in the head. On the other hand, the induced quantities in the central nervous system for the head and the whole body are both similar to each other. Therefore, under these computation conditions, the results suggest that the induced quantities of CNS can be computed by using the head

Table 3. Metrics of ICNIRP and IEEE guidelines at 150 kHz.

| Metrics | | Values | Guidelines |
|--------------------|------------------------------|---------|-------------|
| J (A/m^2) | | 0.3 | ICNIRP1998 |
| E_{99} (V/m) | | 20.25 | ICNIRP2010 |
| E_{5mm} (V/m) | brain | 44.175 | IEEE |
| | heart | 847.006 | |
| | limbs | 94.0299 | |
| | other tissues | 31.3881 | |
| SAR_{10g} (W/kg) | head and trunk | 2 | ICNIRP/IEEE |
| | Limbs (pinnae ^a) | 4 | |
| SAR_{wm} (W/kg) | | 0.08 | |

SAR_{10g} is the localized SAR average of any 10g cubical volume of tissue.

SAR_{wm} is the whole-mass average SAR.

^aIEEE includes pinnae together with limbs for localized SAR.

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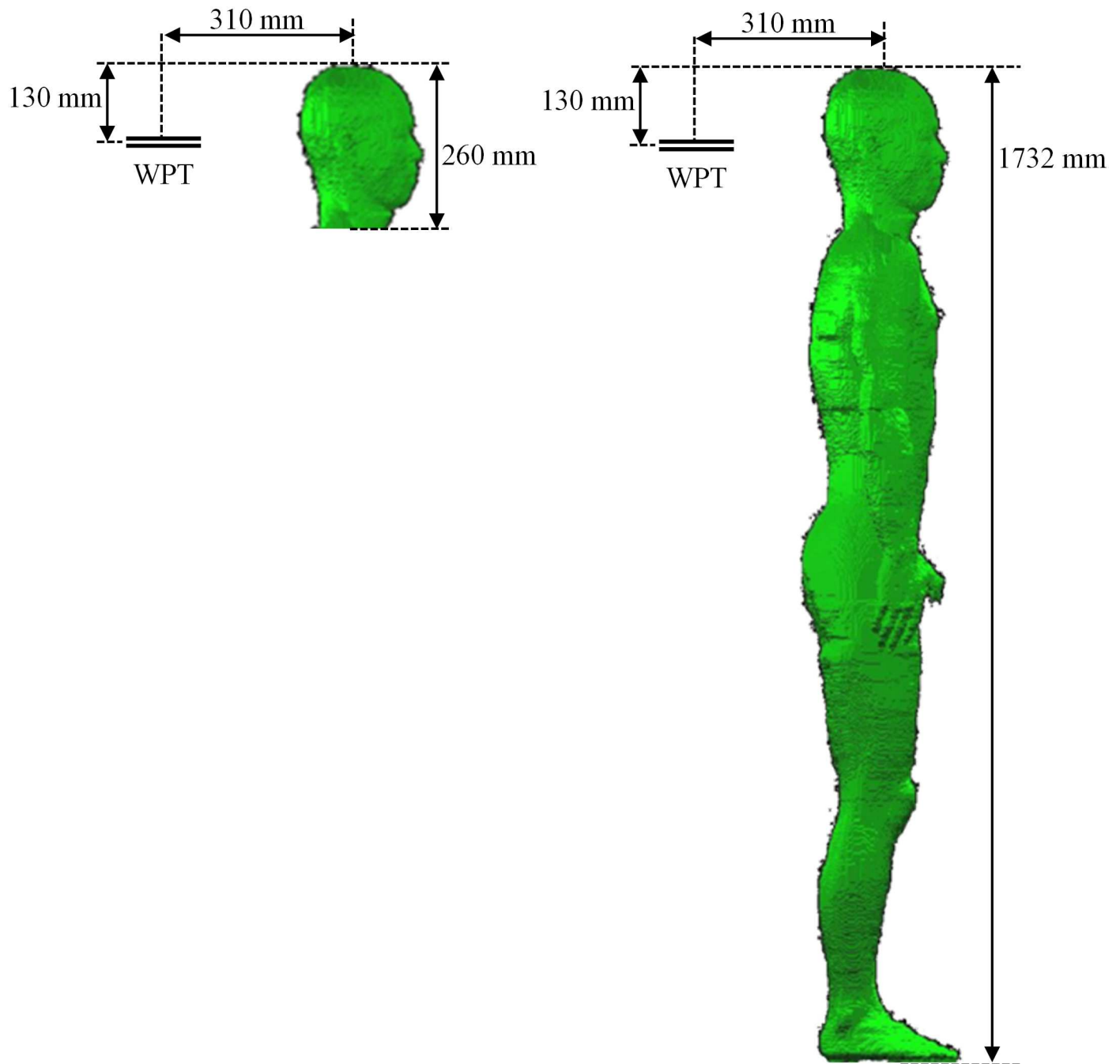


Fig 7. Position of the human body model in reference to a WPT system.

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model instead of the whole-body model, whereas the other induced quantities should be computed by using the whole-body model. We investigated the maximum allowable powers (MAP) to comply with the basic restrictions suggested by each safety guideline as shown in Fig 9. In this computation condition, the results show that the basic restrictions to protect from electrostimulation effects (J , E_{99} , $E_{5\text{mm}}$) are more severe than the basic restrictions to protect from the thermal effects (SAR) for all guidelines, while for the current density (J), the limits

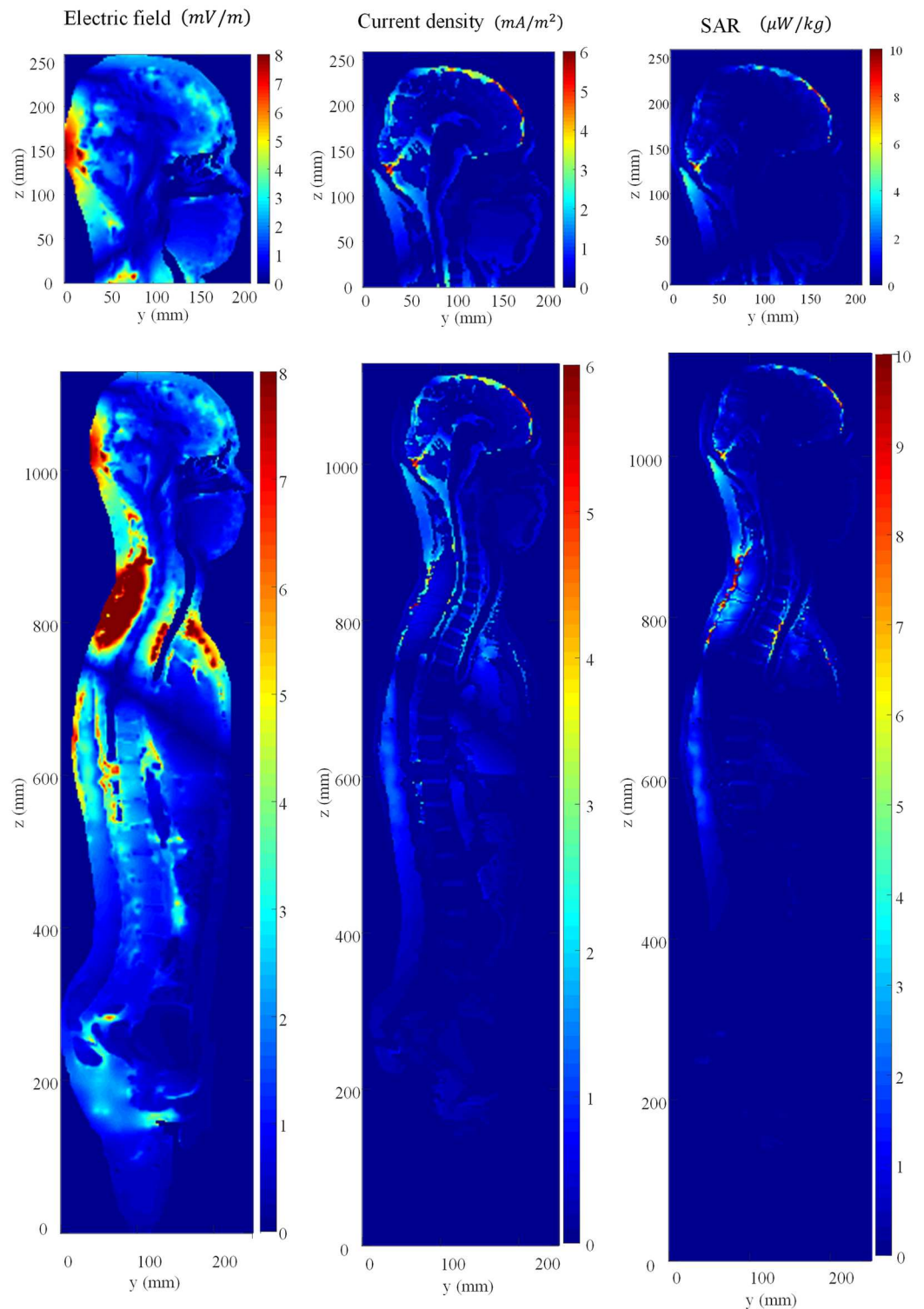


Fig 8. Distributions of induced electric field, current density, and SAR in head and whole-body models in the cross section at xy plane, respectively.

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Table 4. Induced quantities in head and whole body.

| | Head (tissue name) | Whole body (tissue name) |
|--------------------------------|-----------------------|--------------------------|
| J (mA/m ²) | 3.48 (CSF) | 4.15 (Muscle) |
| J_{cns} (mA/m ²) | 1.40 (Grey matter) | 1.41 (Cerebellum) |
| E_{99} (mV/m) | 9.36 (Cortical bone) | 11.57 (Cortical bone) |
| E_{99cns} (mV/m) | 7.61 (CSF) | 7.63 (CSF) |
| E_{5mm} (mV/m) | 19.53 (Cortical bone) | 31.09 (Fat) |
| E_{5mmcns} (mV/m) | 10.93 (Grey matter) | 10.95 (Grey matter) |
| SAR_{1g} (nW/kg) | 6.97 (CSF) | 21.86 (Muscle) |
| SAR_{10g} (nW/kg) | 2.63 (Cerebellum) | 15.65 (Muscle) |
| SAR_{wm} (nW/kg) | 0.44 | 0.30 |

CSF is the abbreviation for Cerebrospinal fluid.

The induced quantities including the cns subscript represent the values in the central nervous system

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Maximum Allowable Power (W)

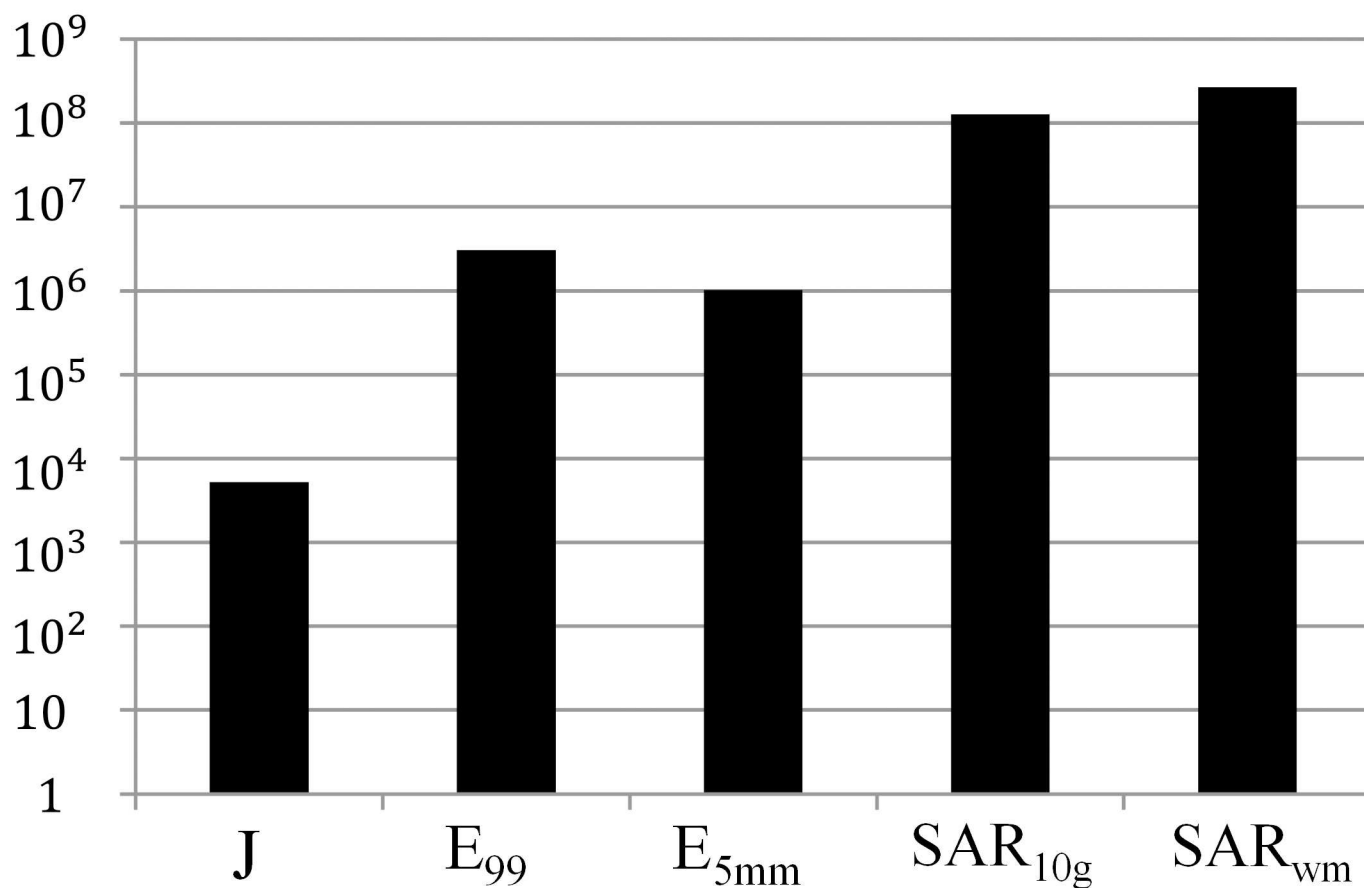


Fig 9. Maximum allowable power to satisfy the basic restrictions.

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recommended by the 1998 ICNIRP are the most severe. Thus, in order to comply with the human exposure regulations for the WPT system operating at 150 kHz frequency, we should first consider the basic restrictions to protect from the effects of electrostimulation.

Conclusions

A two-step approach using the quasi-static FDTD has been provided to compute human exposure assessment for an arbitrary non-uniform field of low frequency range. A verification of our method against a theoretical solution for a lossy sphere as a simplified human phantom in the proximity of a magnetic dipole source attests to its accuracy. The result suggests that the proposed method provides an effective tool for computations of induced electric fields in the human body in close proximity to any sources. For application, induced electric field, current density, and SAR in the head and whole body of a human in the proximity to a WPT system for cell phone charging operating at 150 kHz frequency have been computed with the proposed method. The results showed that the limits of induced electric field and current density should be carefully considered in comparison to SAR limits. The proposed method will be applied for accurate evaluation of fields in the body in various EMF environments, such as household, industry, automobile, and medical uses. For future work, the dosimetry for electric vehicle wireless chargers requiring high power will be conducted.

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Author Contributions

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Methodology: SP.

Software: SP.

Supervision: SP.

Validation: MK.

Writing – original draft: SP.

Writing – review & editing: MK.

References

1. Kurs A, Karalis A, Moffatt R, Joannopoulos JD, Fisher P, Soljacic M. Wireless Power Transfer via Strongly Coupled Magnetic Resonances. *Science* (80-) 2007; 317:83–6.
2. Hui SYR, Zhong W, Lee CK. A Critical Review of Recent Progress in Mid-Range Wireless Power Transfer. *IEEE Trans Power Electron* 2014; 29:4500–11.
3. Hurley W, Kassakian J. Induction heating of circular ferromagnetic plates. *IEEE Trans Magn* 1979; 15:1174–81.
4. Jang Y, Jovanović MM. A contactless electrical energy transmission system for portable-telephone battery chargers. *IEEE Trans Ind Electron* 2003; 50:520–7.
5. Schuder J, Stephenson H, Townsend J. High-level electromagnetic energy transfer through a closed chest wall. *Inst Radio Engrs Int Conv Rec* 1961.

6. Green a. W. 10 kHz inductively coupled power transfer—concept and control. *Proc 5th Int Conf Power Electron Var Drives* 1994;1994:694–9.
7. Kissin MLG, Boys JT, Covic GA. Interphase mutual inductance in polyphase inductive power transfer systems. *IEEE Trans Ind Electron* 2009; 56:2393–400.
8. Elliott G a. J, Covic G a., Kacprzak D, Boys JT. A New Concept: Asymmetrical Pick-Ups for Inductively Coupled Power Transfer Monorail Systems. *IEEE Trans Magn* 2006; 42:3389–91.
9. Choi B, Nho J, Cha H, Ahn T, Choi S. Design and Implementation of Low-Profile Contactless Battery Charger Using Planar Printed Circuit Board Windings as Energy Transfer Device. *IEEE Trans Ind Electron* 2004; 51:140–7.
10. Hui SYR, Ho WWC. A new generation of universal contactless battery charging platform for portable consumer electronic equipment. *IEEE Trans Power Electron* 2005; 20:620–7.
11. Liu X, Hui SY. Simulation study and experimental verification of a universal contactless battery charging platform with localized charging features. *IEEE Trans Power Electron* 2007; 22:2202–10.
12. Wireless Power Consortium Website n.d. <https://www.wirelesspowerconsortium.com/> (accessed June 1, 2016).
13. AirFuel Alliance Website n.d. <http://airfuel.org/> (accessed June 1, 2016).
14. Lin JC. Wireless Power Transfer for Mobile Applications, and Health Effects [Telecommunications Health and Safety]. *IEEE Antennas Propag Mag* 2013; 55:250–3.
15. ICNIRP. Comment on the ICNIRP guidelines for limiting exposure to time-varying electric, magnetic, and electromagnetic fields (up to 300 GHz). *Health Phys* 1998; 74:494–522. PMID: [9525427](#)
16. ICNIRP. Fact Sheet on the Guidelines for Limiting Exposure to Time-Varying Electric and Magnetic Fields (1Hz–100Hz). *Health Phys* 2010; 99:818–36. PMID: [21068601](#)
17. IEEE Standard for Safety Levels with Respect to Human Exposure to Electromagnetic Fields, 0–3 kHz. *IEEE Standard C95.6*; 2002.
18. IEEE Standard for Safety Levels With Respect to Human Exposure to Radio Frequency Electromagnetic Fields, 3 kHz to 300 GHz. *IEEE Standard C95.1*; 2005.
19. Gandhi O, Furse C. Millimeter-resolution MRI-based models of the human body for electromagnetic dosimetry from ELF to microwave frequencies. *Proc Int Work Voxel Phantom Dev* 1995.
20. Dimbylow P. The development of realistic voxel phantoms for electromagnetic field dosimetry. *Proc Int Work Voxel Phantom Dev* 1996.
21. Mason PA, Ziriax JM, Hurt WD, Walters TJ, Ryan KL, Nelson DA, et al. Recent advancements in dosimetry measurements and modeling. *Radio Freq Radiat Dosim Its Relatsh to Biol Eff Electromagn Fields* 2000:141–55.
22. Nagaoka T, Watanabe S, Sakurai K, Kunieda E, Watanabe S, Taki M, et al. Development of realistic high-resolution whole-body voxel models of Japanese adult males and females of average height and weight, and application of models to radio-frequency electromagnetic-field dosimetry. *Phys Med Biol* 2004; 49:1–15. PMID: [14971769](#)
23. Kunz K, Luebbers R. The finite difference time domain method for electromagnetics. 1993.
24. Sun Y, Song H, Jara AJ, Bie R. Internet of Things and Big Data Analytics for Smart and Connected Communities. *IEEE Access* 2016; 4:766–73.
25. Baccarelli E, Cordeschi N, Mei A, Panella M, Shojafar M, Stefa J. Energy-efficient dynamic traffic off-loading and reconfiguration of networked data centers for big data stream mobile computing: Review, challenges, and a case study. *IEEE Netw* 2016; 30:54–61.
26. Shojafar M, Cordeschi N, Amendola D, Baccarelli E. Energy-saving adaptive computing and traffic engineering for real-time-service data centers. 2015 IEEE Int. Conf. Commun. Work. ICCW 2015, Institute of Electrical and Electronics Engineers Inc.; 2015, p. 1800–6.
27. Deford J, Gandhi O. An impedance method to calculate currents induced in biological bodies exposed to quasi-static electromagnetic fields. *Electromagn Compat IEEE*. . . 1985.
28. Dawson TW, De Moerloose J, Stuchly MA. Comparison of magnetically induced ELF fields in humans computed by FDTD and scalar potential FD codes. *Appl Comput Electromagn Soc J* 1996; 11:63–71.
29. Laakso I, Tsuchida S, Hirata A, Kamimura Y. Evaluation of SAR in a human body model due to wireless power transmission in the 10 MHz band. *Phys Med Biol* 2012; 57:4991–5002. doi: [10.1088/0031-9155/57/15/4991](#) PMID: [22801053](#)
30. Christ A, Douglas M, Roman J, Cooper E, Sample A, Waters B, et al. Evaluation of wireless resonant power transfer systems with human electromagnetic exposure limits. *IEEE Trans Electromagn Compat* 2013; 55:265–74.

31. Park SW, Wake K, Watanabe S. Incident electric field effect and numerical dosimetry for a wireless power transfer system using magnetically coupled resonances. *IEEE Trans Microw Theory Tech* 2013; 61:3461–9.
32. Park SW. Dosimetry for Resonance-Based Wireless Power Transfer Charging of Electric Vehicles. *J Electromagn Eng Sci* 2015; 15:129–33.
33. Song H, Shin H, Lee H, Yoon J, Byun J. Induced Current Calculation in Detailed 3-D Adult and Child Model for the Wireless Power Transfer Frequency Range. *IEEE Trans Magn* 2014; 50:3–6.
34. Moerlose JD, Dawson TW, Stuchly MA. Application of the finite difference time domain algorithm to quasi-static field analysis. *Radio Sci* 1997; 32:329–41.
35. Hirata A, Takano Y, Nagai T. Quasi-Static FDTD method for dosimetry in human due to contact current. *IEICE Trans Electron* 2010.
36. Stratton J. *Electromagnetic theory*. 2007.
37. Gabriel C. *Compilation of the Dielectric Properties of Body Tissues at RF and Microwave Frequencies*. Environ Heal 1996; Report No.:21.
38. Harrington R, Harrington J. *Field computation by moment methods*. 1996.