



OPEN

Design and SAR assessment of three compact 5G antenna arrays

A. Lak¹, Z. Adelpour^{1✉}, H. Oraizi² & N. Parhizgar¹

In this paper three different multi stub antenna arrays at 27–29.5 GHz are designed. The proposed antenna arrays consist of eight single elements. The structure of feeding parts is the same but the radiation elements are different. The feeding network for array is an eight way Wilkinson power divider (WPD). To guarantee the simulation results, one of the proposed structures is fabricated and measured (namely the characteristics of S_{11} , E-, and H-plane patterns) which shows acceptable consistency with measurement results. The simulation results by CST and HFSS show reasonable agreement for reflection coefficient and radiation patterns in the E- and H- planes. The overall size of the proposed antenna in maximum case is $29.5 \text{ mm} \times 52 \text{ mm} \times 0.38 \text{ mm}$ ($2.8 \lambda_0 \times 4.86 \lambda_0 \times 0.036 \lambda_0$). Moreover, for Specific Absorption Rate (SAR) estimation, a three-layer spherical human head model (skin, skull, and the brain) is placed next to the arrays as the exposure source. The simulation results show that the performance of proposed antennas as low-SAR sources makes them ideal candidates for the safe usage and lack of impact of millimeter waves (mmW) on the human health. In all three cases of SAR simulations the value of SAR_{1g} and SAR_{10g} are below the standard limitations.

Recently the 5G technology has become an attractive subject in the telecommunication industry. Upcoming 5G systems should satisfy several requirements such as: higher bandwidth, low latency, broad coverage of network, high reliability, high throughput, high connection density, low power consumption, high gain¹. Some frequency bands have been proposed as candidates for millimeter wave (mmW) for example 27–29.5 GHz, 36–40 GHz, 47.2–50.2 GHz². High path loss owing to reduced size of antenna dimensions and increasing atmospheric absorption are two problems at high frequency. Although higher data rates can support by these frequency bands but the signal wavelength becomes shorter and according to the Friis equation, the free space path loss becomes higher^{3,4}. Employment of high gain directive antennas or antenna array is a solution to compensate such problems, which provides multipath suppression and interference mitigation however low radiation toward human tissues is expected to achieve low specific absorption rate⁵.

The 5G antennas usually use in handheld devices, such as tablets and mobile phone therefore they evidently should be small in size and light weight. It has been demonstrated when the radiation pattern of antenna is directed to the top or bottom edges of the devices (that is endfire pattern) the influence of user's hand on the antenna radiation is minimize⁶. Antenna arrays at 5G systems can be designed by some technologies such as microstrip and SIW^{7,8}, and in many types like fermi, vivaldi, quasi yagi, and cavity backed^{9–11}. However, the effects of electromagnetic field on human body tissue should evaluate by possible methods like numerical methods to ensure that these field sources do not threaten human health at 5G frequency bands. To appraise the exposure some parameters use by standard institutes such as Specific Absorption Rate (SAR), power density (PD), and the Skin Surface Temperature Elevation. There are some standards, such as Federal Communications Commission (FCC) and IEEE to determine the permissible values of SAR from exposure to electromagnetic fields for human safety. Their values are different for occupational and public environments. According to these standards the SAR_{1g} and SAR_{10g} limits are 1.6 W/kg and 2 W/kg respectively¹².

The studies about SAR levels on human tissues have been done in many various conditions and methods such as in vivo–in vitro environment and also by numerical methods. Due to the probable hazards on human health in actual conditions, many assessments about field exposure are conducted by software simulations and experimental environments. In¹³, for the determination of SAR, the human body tissues are modeled in one (skin) and three layer (including skin, fat, and muscle) and a four-element array of rectangular patch antenna as an exposure source have been modeled by the CST software. The input powers were 20 dBm and 24 dBm and the frequencies were 28, 40 and 60 GHz. The results showed that at both power, SAR_{1g} and point SAR values at 28 GHz were lower

¹Department of Electrical Engineering, Shiraz Branch, Islamic Azad University, Shiraz, Iran. ²School of Electrical Engineering, Iran University of Science and Technology, Tehran, Iran. ✉email: adelpour@aut.ac.ir

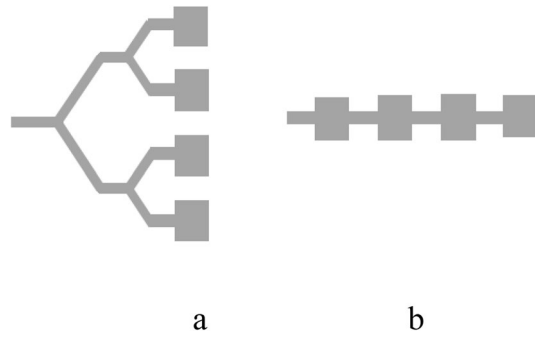


Figure 1. (a) Corporate and (b) series feed array structure¹⁷.

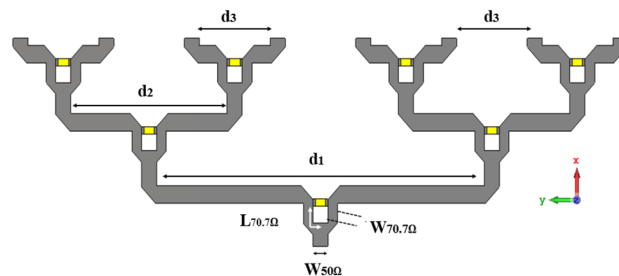


Figure 2. Eight way Wilkinson power divider.

than other frequencies¹⁴, the penetration of radiation at 30 GHz in human ear canal and tympanic membrane have been investigated and the results showed a very low penetration and not notable significant thermal effect on the tympanic membrane. In¹⁵ the absorption of RF field at 39 GHz both in *in vivo* bovine the brain tissue and a brain simulating gel model have been investigated. The results represented the SAR and radiation penetration in the brain model, and therefore SAR, decreases with increasing depth and frequency. In¹⁶ the SAR values in head model of children and adults at 28 GHz (30 mW) and a microstrip antenna as a field source have been simulated. The results showed that absorption in tissues decreasing rapidly in depth. As well as duo to epidermis and dermis thickness (0.1 and 2 mm), the mmW values values are quickly absorbed in these layers and do not reach the deeper tissues. In this paper three compact, lightweight, high gain eight arrays antenna are simulated at 27–29.5 GHz. Design procedure, simulation, and measurement results are presented in the following sections. Also the SAR_{1g} and SAR_{10g} have been simulated and evaluated to determine the specific absorption rate.

Antenna design

Feeding part. There are different types of feed network for feeding an array antenna. The formal array feeding networks are series or corporate feed network based on microstrip structures¹⁷, as shown in Fig. 1.

Microstrip array has a simple structure and easy fabrication process, which leads to compact and low-cost structures, but duo to its high losses in mmW frequency band the endfire antenna is recommended to use^{17,18}. Some types of passive power divider networks are Wilkinson, T-junction, and Resistive power divider. T-junction is lossless but it has two disadvantages: un-matched at all ports and no isolation between output ports. The resistive type can be matched at all ports but it is lossy and doesn't have isolation between output ports. But Wilkinson is lossless (if all ports are matched) and has good isolation.

In this paper Wilkinson Power Divider (WPD) has been adopted. To evaluate the WPD performance three parameters should be checked: reflection coefficients, coupling and isolation between ports¹⁸. In two-way WPD, the isolation resistor is $2Z_0$ and the impedance of $\lambda/4$ is $\sqrt{2}Z_0$. For equal WPD (or 3 dB) the $Z_0 = 50\Omega$, the impedance of $\lambda/4$ is $\sqrt{2}Z_0 = 70.7\Omega$ and isolation resistor is $2Z_0 = 100\Omega$ ¹⁸. To design WPD at 28 GHz the TXline calculator is used. The substrate is Rogers RT/Duriod 5880 with 0.38 mm thickness, loss tangent of 0.0009 and relative permittivity of $\epsilon_r = 2.2$. The values for WPD are obtained as: $W_{50\Omega} = 1.18$ mm, $W_{70.7\Omega} = 0.65$ mm and $L_{70.7\Omega} = 1.97$ mm (Fig. 2). The isolation resistor is 100Ω (size is 1×0.5 mm²) from 0402 SMD family. For eight-way WPD, three stages of two ways WPD is needed. As shown in Fig. 2, d_1 and d_2 are approximately 4 times and 2 times longer than d_3 , respectively. The distance between two output ports (d_3) is about $\frac{\lambda}{2}$ to satisfy the array considerations. The performance of the eight-way designed WPD has been shown in Fig. 3. As it can be seen reflection coefficient, isolation and insertion loss are in acceptable range and the observed deviation from the theoretical values are due to high frequency range of operation which leads to higher microstrip line loss (conductor, dielectric and radiation losses)^{18–20}.

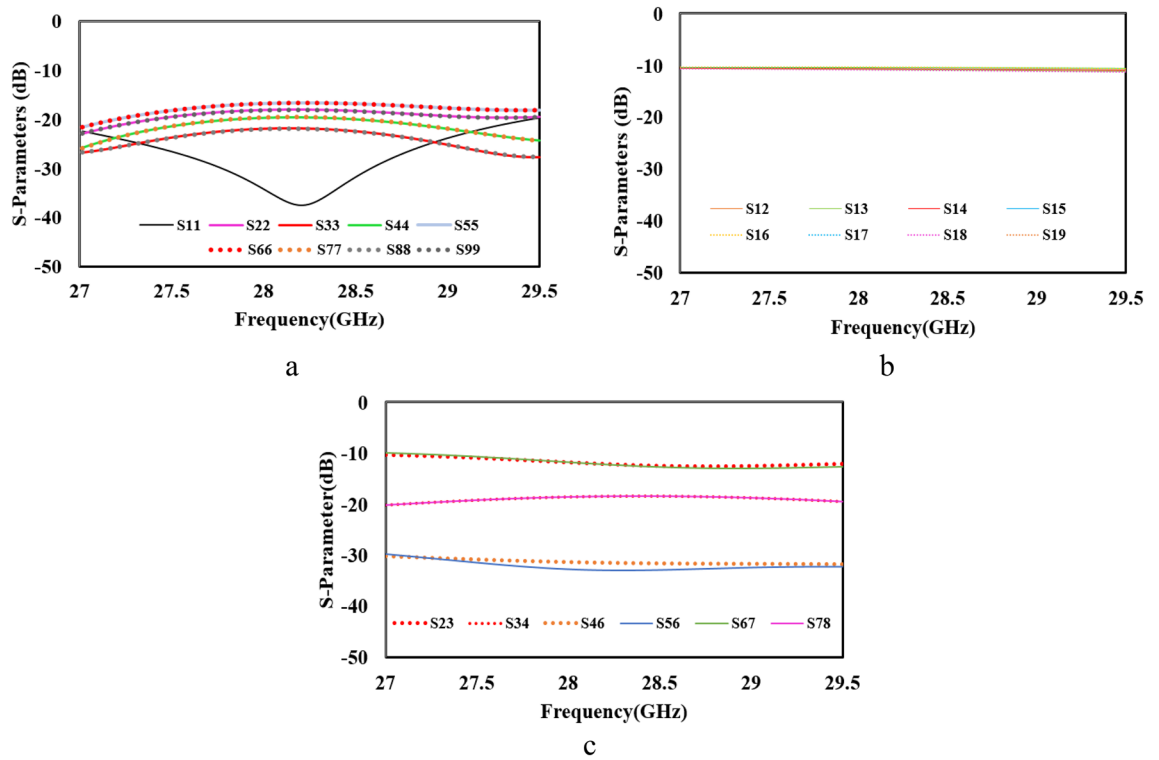


Figure 3. Simulation results of desinged eight-way WPD (a) reflection coefficients, (b) insertion loss and (c) isolation.

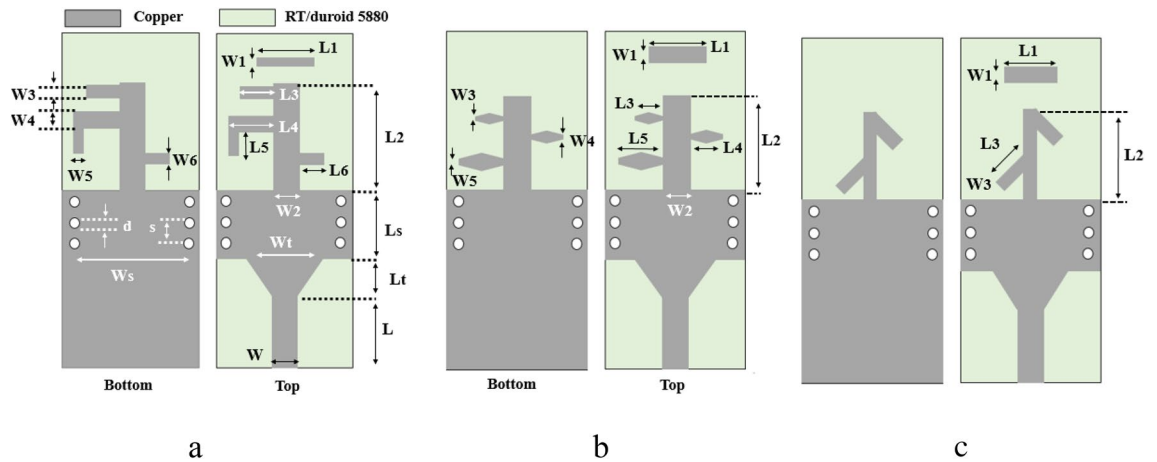


Figure 4. Structure of the proposed single element (a) antenna1, (b) antenna2 and (c) antenna3²¹.

Single elements. The design procedure of three different single elements is completely described in²¹. Figure 4 shows the structures. The substrate is RT/duriod 5880 with 15mil thickness, $\epsilon_r = 2.2$ and $\tan\delta = 0.0009$. The dimension of antenna1 according to Fig. 4a are $L_1 = 2.5$ mm, $L_2 = 5$ mm, $L_3 = 1.615$ mm, $L_4 = 2.275$ mm, $L_5 = 1.25$ mm, $L_6 = 1.125$ mm, $W_1 = 0.4$ mm, $W_2 = 1.2$ mm, $W_3 = 0.5$ mm, $W_4 = 0.75$ mm, $W_5 = 0.5$ mm, $W_6 = 0.5$ mm. For antenna2 the dimensions in Fig. 4b are $L_1 = 2.8$ mm, $L_2 = 4.22$ mm, $L_3 = 1.068$ mm, $L_4 = 1.425$ mm, $L_5 = 1.9$ mm, $W_1 = 0.75$ mm, $W_2 = 1.25$ mm, $W_3 = 0.14$ mm, $W_4 = 0.1875$ mm, $W_5 = 0.25$ mm. In addition, for the last one in Fig. 4c the dimensions are $L_1 = 2.5$ mm, $L_2 = 3.9$ mm, $L_3 = 1.8$ mm, $W_1 = 0.5$ mm, $W_2 = 0.6$ mm, $W_3 = 0.6$ mm. For the feeding part (which is the same for all antenna) the calculated parameters are $L_s = 3.5$ mm, $w_t = 3.2$ mm, $L = 3.5$ mm, $w_1 = 1.2$ mm, $w = 5.5$ mm, $L_t = 1.6$ mm, $d = 0.6$ mm, $s = 1.2$ mm. The details of the design procedure for each of the single elements and the results (simulation and fabrication) of them are reported in²¹. All of these antennas have end-fire patterns and acceptable measurmant performance but are not applicable in 5G systems due to low gain values as the single element. Moreover, regarding the Ferris equation, the path loss become higher as the frequency increases. Accordingly, to overcome the path loss in 5G mobile communication system,

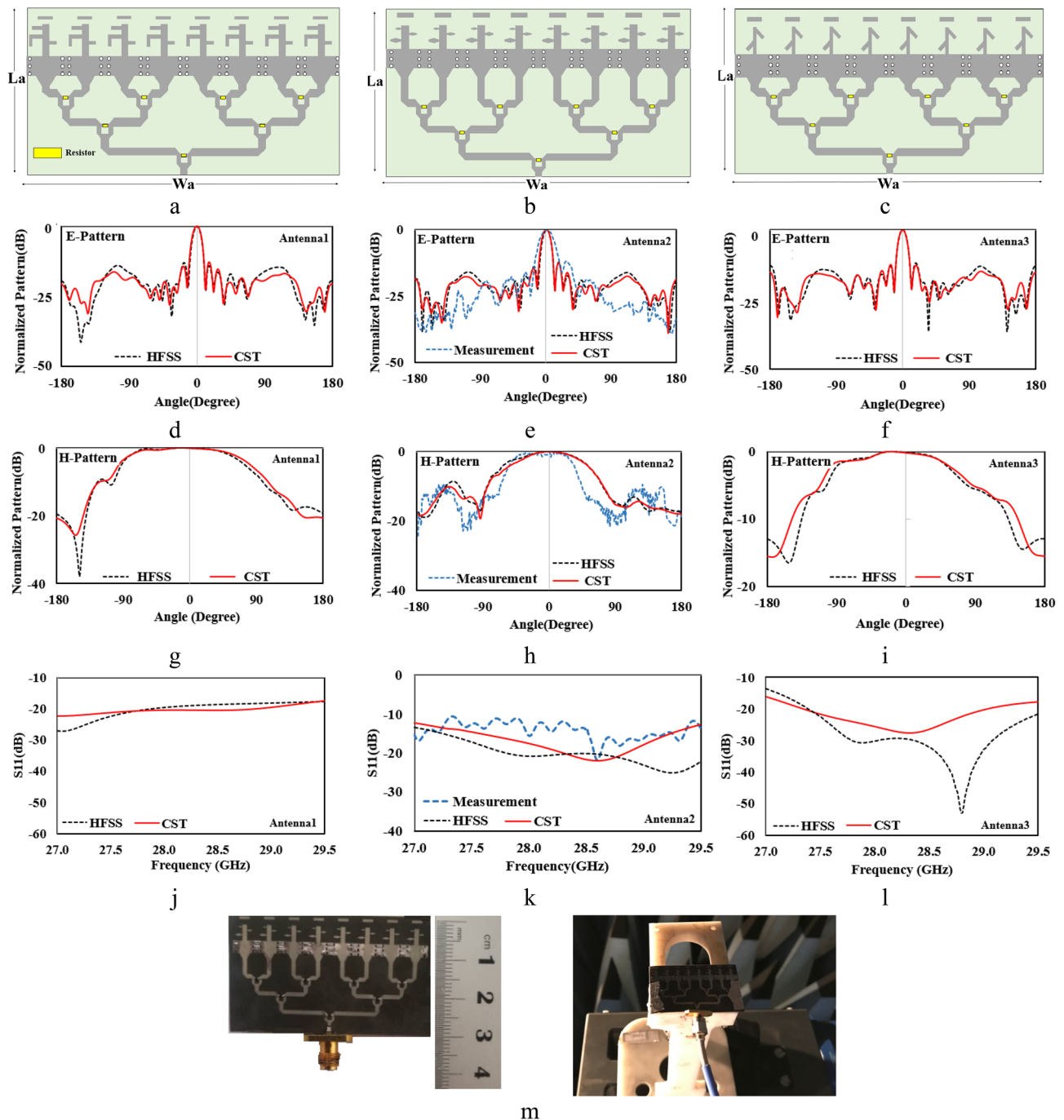


Figure 5. The array structure and simulation results of antenna1 (a,d,g,j), antenna2 with measurement results (b,e,h,k) and antenna3 (c,f,i,l) and (m) the prototype of proposed antenna2 with SMK connector.

minimum value of 12 dB gain is required²². So, the antenna array configuration is proposed to achieve the required gain value.

Linear array antennas. Generally, the number of antenna array elements are 2^N owing to 2^N -way is beneficial structure for designing a power divider with minimum losses. In addition, impedance matching can be accomplished easily²³. The schematics of three different array antennas are shown in Fig. 5a–c. For better evaluation of array performance, two full-wave softwares (CST and HFSS) is used for simulation and the results of each array is shown in Fig. 5, respectively.

As it can be observed, there is a good consistency between the simulation results in both softwares. Between these three proposed antennas, antenna2 is chosen to be fabricated and tested as shown in Fig. 5m. The connector that is used is SMK with frequency range up to 40 GHz. So, in Fig. 5, the measurement results are shown for antenna2, too. As it can be seen, the measurement and simulation results for antenna2 are in good agreement,

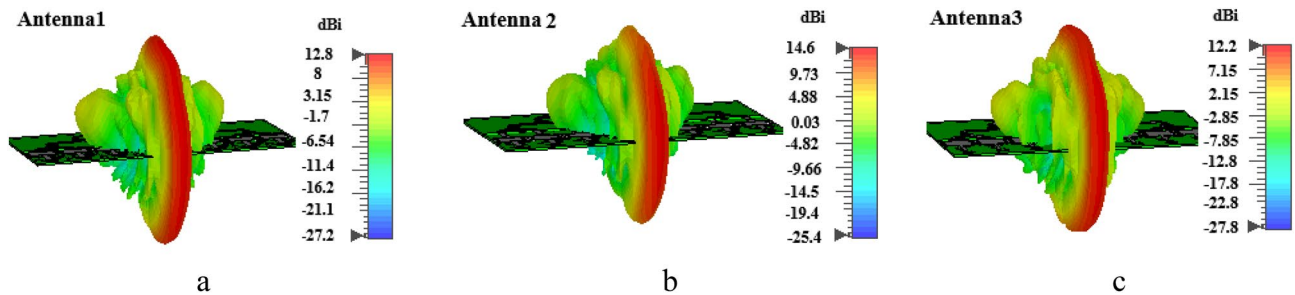


Figure 6. The 3D radiation pattern and calculated gain value of: (a) antenna1, (b) antenna2 and (c) antenna3.

Parameter	Size (mm ³)
Antenna 1	26.42 × 53.6 × 0.38
Antenna 2	29.52 × 52 × 0.38
Antenna 3	26.42 × 52 × 0.38

Table 1. The size of the proposed antennas.

which suggests that the other two antennas are also applicable in this frequency band. The differences between measurement and simulation results can be considered due to substrate and specially connector losses. It is obvious that performance in high frequency range duplicates the radiation and thermal loss effect of soldering and SMD resistors. Moreover the loss and errors of fabrication and measurement devices can not be omitted. In Fig. 6, the simulated gain values are shown which are high enough for handheld 5G systems. Moreover, the endfire pattern of the proposed structures is suitable for 5G frequency bands because of its capability (of array antenna) to compensate the path loss. As will be discussed in next section, directive antenna is a solution to minimize the SAR values in human tissue.

The size of three proposed structures are presented in Table 1 which shows acceptable reflection coefficient and gain values while keeping the overall size as minimum as possible as a good candidates for handheld 5G systems. The gain values are 12.85 dB, 14.6 dB, and 12.2 dB for antenna1, antenna2, and antenna3 respectively.

SAR assessments of proposed array antennas

The 5G systems have many interesting advantages, such as higher bandwidth and data rate; hence, they are growing surprisingly in the world. However, their probably adverse effects on human body tissues from such electromagnetic sources should appraise to ensure safety of human body. Some biological effects of electromagnetic fields such as cancer, blood brain barrier, the brain tumor, Cataract, skin disease, sleep disorder have been reported^{24–27}. The other effects of mmW frequency are genotoxicity (DNA damage), cell proliferation, gene expression, cell signaling, electrical activity, and membrane effects have been briefed in²⁸. Many references can be cited in which the advantages and disadvantages of SAR and PD parameters are discussed. Some of these references prefer SAR and the others PD. It seems that this issue is still an open subject, which needs to be investigated more carefully. Owing to the following reasons in this paper, the SAR is chosen. According to FCC, the power density (PD) unit is used for the distances of 5 cm or more. Therefore, it only deals with far-field exposures and does not consider the near field exposures. On the other hand, some of the mmW devices such as handsets or tablets use almost near to the head, hand or in the pocket next to the human body (in a few millimeters distances i.e. near field region) and in these conditions, the PD is not a suitable unit to evaluate the human safety. In addition, estimations based on PD do not describe the absorbed power and distributed field, but only exhibit the travelling wave in human tissues. Hence, the SAR technique is used to study^{13,29–32}.

There are some limitations to assess the SAR value in human body, because the adverse biological effects may occur, so the numerical simulations are to be used for SAR evaluation. SAR is a unit to determine the rate of how much energy from electromagnetic source is absorbed per mass unit by human tissues as shown in Eq. (1).

$$SAR = \sigma \frac{E_i^2}{\rho} (W/Kg) \quad (1)$$

where σ is the conductivity of tissue in unit (S/m), E is the electric field intensity in unit (V/m), ρ is the mass density of tissue in unit (kg/m³). The SAR averages either over the whole body, or over a small sample volume (typically 1 g or 10 g of tissue). The unit of SAR is watt per kilogram²¹. SAR limits in International Commission on Non-Ionizing Radiation Protection (ICNIRP) and the IEEE C95.1–2019 standards is 2 W/kg over 10 g and according to FCC standard SAR limit for 1 g is 1.6 W/kg. These limits are for the frequencies up to 10 GHz and 6 GHz respectively. The SAR limits above these two frequencies for near field exposures at mmW have not been proposed yet which is due to near field exposure at mmW. However, it is an important topic to study.

Tissues	ϵ_r	σ (S/m)	ρ (kg/m ³)	Radius (mm)
Skin	18.71	26.19	1100	106.5
Skull	7.51	8.88	1990	105.1
Brain	18.59	21.86	1041	98.3

Table 2. Human head model properties at 28 GHz.

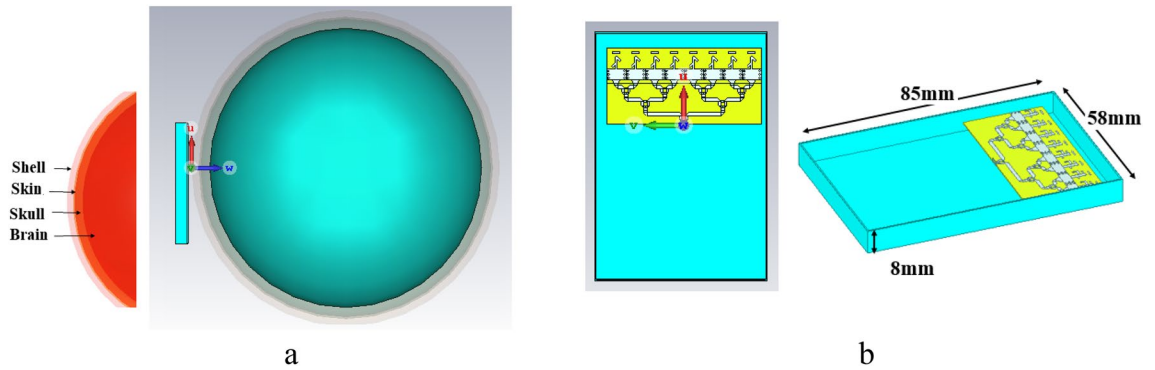


Figure 7. (a) Human head and handset model. (b) The handset dimension.

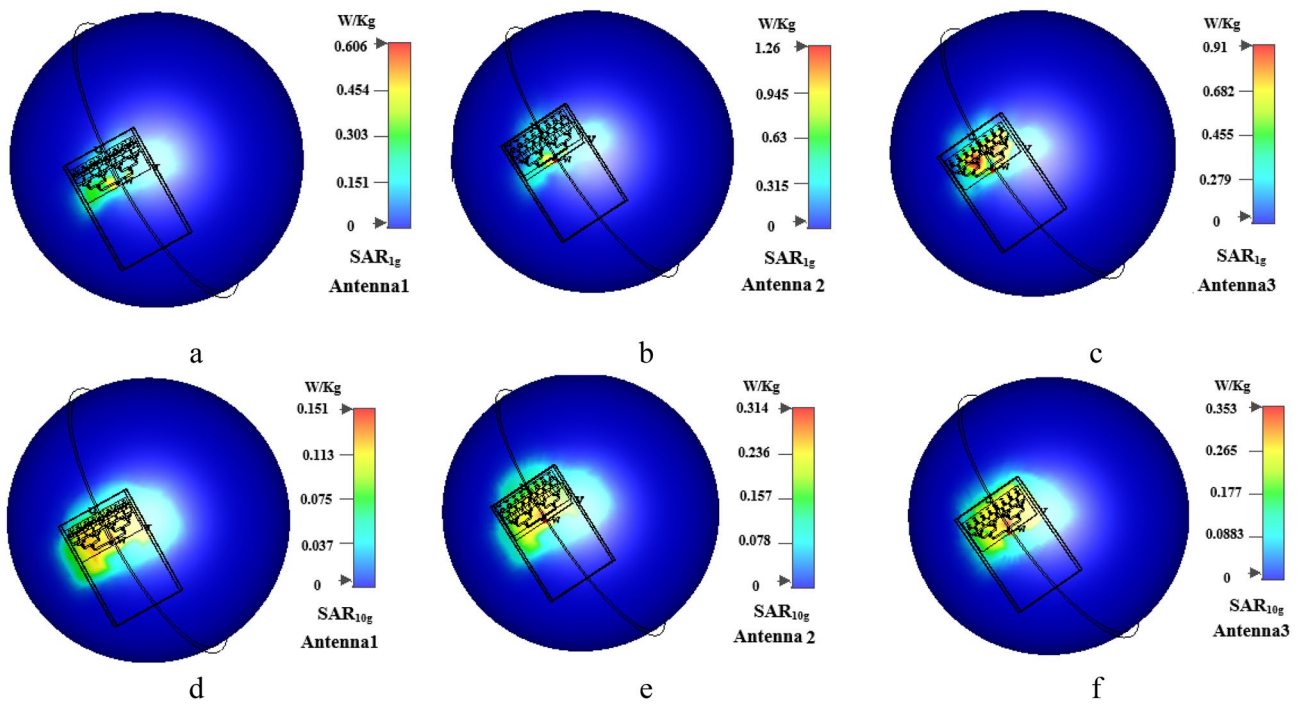


Figure 8. Simulated SAR parameter of antenna1 (SAR_{1g} (a), SAR_{10g} (d)), antenna2 (SAR_{1g} (b), SAR_{10g} (e)) and antenna3 (SAR_{1g} (c), SAR_{10g} (f)).

Head and handset model. To simulate the SAR parameter, a three-layer spherical human head model including skin, skull, and the brain is situated near the antenna as an exposure source. All human tissues have different permittivity (ϵ_r) and conductivity (σ), and their properties depend on many parameters such as frequency, age, etc. At 28 GHz, the properties and radius of three layers are listed in Table 2^{3,21}. The covering shell is a low loss dielectric with relative permittivity of $\epsilon_r = 4.5$. The human head and handset model are shown in Fig. 7.

For the handset a plastic housing box with $58 \times 85 \times 8$ mm³ dimension is used with $\epsilon_r = 3$ and $\sigma = 0.02$ S/m and 1 mm thickness in which the proposed antenna array is placed on top³³. A glass with $\epsilon_r = 5.5$ used too as a screen of the handset⁸. The input power for the antennas in 5G systems can be set to 15 dBm, 18 dBm and 20 dBm according to FCC³⁴ and the distance between head and antenna are 5 mm^{3,13,34}. Figure 8. shows the results

Antenna array	SAR (W/kg)	Dipole antenna	SAR (W/kg)	Single element	SAR (W/kg)	Dipole antenna	SAR (W/kg)
Ant.1(SAR1g)	0.606	SAR _{1g}	1.81	Ant.1(SAR1g)	0.236	SAR _{1g}	1.81
Ant.2(SAR1g)	1.26	–	–	Ant.2 (SAR1g)	0.284	–	–
Ant.3(SAR1g)	0.91	–	–	Ant.3 (SAR1g)	0.416	–	–
Ant.1(SAR10g)	0.151	SAR _{10g}	0.51	Ant.1 (SAR10g)	0.081	SAR _{10g}	0.51
Ant.2(SAR10g)	0.31	–	–	Ant.2 (SAR10g)	0.071	–	–
Ant.3(SAR10g)	0.353	–	–	Ant.3 (SAR10g)	0.169	–	–

Table 3. The averaged SAR_{1g} and SAR_{10g} for 15 dBm input power.

Array antenna	SAR (W/kg)	Dipole antenna	SAR (W/kg)
Ant.1 (SAR10g)	0.478	SAR10g	1.62
Ant.2 (SAR10g)	0.998		–
Ant.3 (SAR10g)	0.92		–

Table 4. The averaged SAR_{10g} for 20 dBm input power.

Ref	d (mm)	Power (dBm)	Array	Gain (dB)	SAR (1 g)	SAR (10 g)
¹³	5 mm	24	2 × 2	11.23	1.35	–
³⁵	N.A	N.A.	–	9.485	1.42	0.3
²²	N.A	N.A.	2 × 2	12.3	–	0.37
³⁶	N.A	24	MIMO	≈ 10	–	Max:1.2
			8 × 8	–	–	Min:0.8
Ant. 1	5 mm	15	1 × 8	12.85	0.56	0.15
Ant. 2	5 mm	15	1 × 8	14.6	1.25	0.31
Ant.3	5 mm	15	1 × 8	12.2	0.91	0.35

Table 5. Comparison the SAR of three antenna performances with other 5G antenna references at 28 GHz. d: Distance between antenna and human head model.

of SAR_{1g} and SAR_{10g} for 15 dBm. From this figure, it can be observed that: (1) The SAR at the nearest distance from antennas are more than others are. (2) The SAR_{1g} is higher than SAR_{10g}. (3) By increasing the distance between antenna and human head model the SAR is decreased.

Table 3 shows the simulation results of SAR_{1g} and SAR_{10g} of three proposed array and single antennas with 15 dBm power. The single element SAR values in our published paper²¹ are used in Table 3. As can be observed, all of the simulated values are under the FCC and ICNIRP standard limits for array antennas too. Considering the same feeding part for three proposed structures and different radiating element for each, the antennas have different electric field strength (E) which leads to different SAR values according to Eq. (1).

Although all results for antenna array are larger than single elements. It can be cause by more radiation elements in array type. In the commercial SAR measurement system, a dipole antenna is used to measure the SAR parameter. For better comparison, the SAR values of these antennas i.e. proposed antennas, and dipole antenna (from²¹) at 28 GHz are shown in Table 3. The results of both array and single element SAR values are lower than dipole antenna, which shows that proposed end-fire (directive) antennas have lower SAR rather than common dipole antenna. In Table 4 the simulation results with 20 dBm input power are also presented. It can be seen that the values are lower than standard limits and dipole antenna too.

It is well known that electromagnetic fields can damage human tissues, thus designing the low SAR antenna is desirable for mobile devices such as handsets, which use in human body vicinity to reduce probable adverse health effects. In fact, by decreasing the SAR, the field penetration in the human tissues will decrease.

To compare the results of the SAR values and performance of three array antennas Tables 5 and 6 are provided. From the Table 5 the SAR values for proposed antennas are almost lower than the other works at 28 GHz and all of them are low SAR. From the Table 6 the proposed antennas are smaller than other references and all of them have enough gain for 5G systems.

Conclusion

In this paper three compact, small size, low weight, and low SAR array antennas are designed. The feeding part of them is WPD. Owing to their good patterns and reflection coefficient at 27–29.5 GHz from simulation in CST and HFSS and the measurement data, they are suitable for applying in 5G systems. Since the human health effects

Ref	Antenna type	Substrate	BW (dB)	Relative size (λ_0^3)	Array	Gain (dB)	Size reduction
²	Phase array	Rogers 5880	28	N.A	1 × 8	11	–
⁷	Quasi Yagi	N.A	31–34	3.97 × 0.51 × 0.06	1 × 8	≈ 15.5	35.7%
⁹	Vivaldi	Rogers 5880	24.5–28.5	5.61 × 2.8 × 0.047	1 × 8	11.2	56.57%
³⁷	Vivaldi	Rogers RO4003	25–40	12.61 × 9.1 × 0.02	1 × 4	16	78.97%
³⁸	Quasi-Yagi	Arlon Ad 350	25–27	8.41 × 3.73 × 0.35	1 × 8	10.5–12	95.6%
²³	Vivaldi	ISOLA IS300MD	27.5–28.5	5.6 × 12.6 × 0.074	1 × 4	8.01	90%
³⁹	5G	4 substrate	28	7.55 × 7.55 × 0.22	4 subarray	11.62	96.21%
*	Antenna 1	RT/Duroid 5880	27–29.5	2.46 × 5 × 0.036	1 × 8	12.85	–
*	Antenna 2	RT/Duroid 5880	27–29.5	2.8 × 4.86 × 0.036	1 × 8	14.6	–
*	Antenna 3	RT/Duroid 5880	27–29.5	2.5 × 4.86 × 0.036	1 × 8	12.14	–

Table 6. Comparison of three antenna results with other references of 5G.

from electromagnetic fields is very important subject and the user are worry about it, the SAR_{1g} and SAR_{10g} of the antennas at 15 dBm (and SAR_{10g} at 20 dBm) have been simulated in human head model. All the results are lower the standard limits.

The distance between antenna and human head model is 5 mm. Although using hands-free increase the distance and can reduce the SAR. To more examination, results of SAR are compared with dipole antenna that use in commercial SAR measurement system. One of the methods to reduce the SAR is using of directive antenna⁴⁰. Since our proposed antennas are end fire (directive antennas), thus, the results of SAR are suitable. It is noted that in real SAR measurement systems it is impossible to model the human head model in the layers, because the human tissue equivalent material are in jelly or liquid form. Therefore, it may be said that the commercial results are not accurate and more investigation for better tissue model is necessary.

Received: 1 August 2021; Accepted: 15 October 2021

Published online: 28 October 2021

References

1. IEEE 5G and beyond technology roadmap, White paper, IEEE Future Networks (2017).
2. Ojaroudiparchin, N., Shen M. & Pedersen G. F. Multi-layer 5G mobile phone antenna for multi-user MIMO communications. In *IEEE 23th Telecommunications Forum, TELFOR* (2015).
3. He, W., Xu, B., Gustafsson, M., Ying, Z. & He, S. RF compliance study of temperature elevation in human head model around 28 GHz for 5G user equipment application: Simulation analysis, special section on recent advances on radio access and security methods in 5G networks. *IEEE Access* **6**, 830–838 (2017).
4. Paola, C. D., S. Zhang, K. Zhao, Z. Yning, T. Bolin, and G. F. Pedersen. "Wideband beam-steerable array with hybrid high gain antennas for 5G mobile devices." *IEEE Trans. Antennas Propag* <https://doi.org/10.1109/TAP.2019.2925189> (2019).
5. Dadgarpour, A., Zarghooni, B., Virdee, B. S. & Denidni, T. A. Improvement of gain and evaluation tilt-angle using metamaterial loading for millimeter wave applications. *IEEE Antennas Wirel. Propag. Lett.* **15**, 418–420 (2015).
6. Ruan, X. & Chen, C. H. An end-fire circularly-polarized complementary antenna array for 5G application. *IEEE Trans. Antennas Propag.* **68**(1), 266–274 (2019).
7. Puskely, J., Mikulasek, T. & Raida, Z. Design of a compact wideband antenna array for microwave imaging applications. *Radio Eng.* **22**(4), 1224–1232 (2013).
8. Briqech, Z., Sebak, A. R. & Denidni, T. A. Low-cost wideband mm-wave phased array using the piezoelectric transducer for 5G applications. *IEEE Trans. Antennas Propag.* **65**(12), 6403–6412 (2017).
9. Zhu, Sh., Liu, H., Chen, Zh. & Wen, P. A compact gain-enhanced vivaldi antenna array with suppressed mutual coupling for 5G mmWave application. *IEEE Antennas Wirel. Propag. Lett.* **17**(5), 776–779 (2018).
10. Hwang, I. J., Ahn, B. K., Chae, SCh., Yu, J. W. & Lee, W. W. Quasi-Yagi antenna array with modified folded dipole driver for mmWave 5G cellular devices. *IEEE Antennas Wirel. Propag. Lett.* **18**(5), 971–975 (2019).
11. Chen, Rui-Sen, Sai-Wai Wong, Guan-Long Huang, Yejun He, and Lei Zhu. "Bandwidth-Enhanced High-Gain Full-Metal Filtering Slot Antenna Array using TE101 and TE301 Cavity Modes." *IEEE Antennas and Wireless Propagation Letters* <https://doi.org/10.1109/LAWP.2021.3100919> (2021).
12. Khan, R., Al-Hadi, A. & Soh, P. J. Recent advancements in user effect mitigation for mobile terminal antennas: A review. *IEEE Trans. Electromagn. Compat.* **61**(1), 279–287 (2019).
13. Hamed, T. & Maqsood, M. SAR calculation and temperature response of human body exposure to electromagnetic radiations at 28, 40 and 60 GHz mmWave frequencies. *Prog. Electromagn. Res. M* **73**, 47–59 (2018).
14. Vilagosh, Z., Lajevardipour, A. & Wood, A. Computer simulation study of the penetration of pulsed 30, 60 and 90 GHz radiation into the human ear. *Sci. Rep.* **10**(1), 1–10 (2020).
15. Gultekin, D. H. & Siegel, P. H. Absorption of 5G radiation in brain tissue as a function of frequency. *Power Time IEEE Access* **8**, 115593–115612 (2020).
16. Morelli, M. S., Gallucci, S., Siervo, B. & Hartwig, V. Numerical analysis of electromagnetic field exposure from 5G mobile communications at 28 GHz in adults and children users for real-world exposure scenarios. *Int. J. Environ. Res. Public Health* **18**(3), 107 (2021).
17. Gupta, S. H. Analysis and Design of Substrate Integrated Waveguide-based Antennas for Millimeter Wave, Master's thesis, Department of Electrical and Computer Engineering, Concordia University, Canada, <https://spectrum.library.concordia.ca/981274/> (2016).
18. Pozar, D. M. *Microwave Engineering* 4th edn. (Wiley, 2011).
19. Pozar, D. M. & Schaubert, D. *Microstrip Antennas: The Analysis and Design of Microstrip Antennas and Arrays* (Wiley, 1995).
20. <http://www.everythingRF.com>.

21. Lak, A., Adelpour, Z., Oraizi, H. & Parhizgar, N. Three configurations of compact planar multi-stub microstrip antennas for mmW mobile application. *Int. J. Antenna Propag.* <https://doi.org/10.1155/2021/8848218>. (2021).
22. Khan, J., Sehrai, D. A. & Ali, U. Design of dual band 5G antenna array with SAR analysis for future mobile handsets. *J. Electr. Eng. Technol.* **14**(2), 809–816 (2018).
23. Lee, W. W., Hwang, I. J. & Jang, B. End-fire Vivaldi antenna array with wide fan-beam for 5G mobile handsets. *IEEE Access* **8**, 118299–118304 (2020).
24. Lak, A. & Oraizi, H. Evaluation of SAR distribution in six-layer human head model. *Int. J. Antennas Propag.* **12**(1), 56–64. <https://doi.org/10.1155/2013/580872> (2013).
25. Lak, A. Human health effects from radiofrequency and microwave fields. *J. Basic Appl. Sci. Res.* **2**(12), 12302–12305 (2012).
26. Ramundo-Orlando, A. Effects of millimeter waves radiation on cell membrane—A brief review. *J. Infrared Millim. Terahertz Waves* **31**(12), 1400–1411 (2010).
27. Wang, H.-Y. *et al.* The specific absorption rate in different brain regions of rats exposed to electromagnetic plane waves. *Sci. Rep.* **9**(1), 1–13 (2019).
28. Karipidis, K., Mate, R., Urban, D., Tinker, R. & Wood, A. 5G mobile networks and health—A state-of-the-science review of the research into low-level RF fields above 6 GHz. *J. Exposure Sci. Environ. Epidemiol.* <https://doi.org/10.1038/s41370-021-00297-6> (2021).
29. Wu, T., Rappaport, T. S. & Collins, C.M. The human body and millimeter wave wireless communication systems: interactions and implications. In *IEEE International Conference on Communications, London, UK* (2015).
30. Chahat, N., Zhadobov, M., Le Coq, L., Alekseev, S. & Sauleau, R. Characterization of the Interactions between a 60-GHz antenna and the human body in an off-body scenario. *IEEE Trans. Antennas Propag.* **60**, 5958–5965 (2012).
31. Colombi, D., Thors, B. & Törnevik, C. Implications of EMF exposure limits on output power levels for 5G devices above 6 GHz. *IEEE Antennas Wirel. Propag. Lett.* **14**, 1247–1249 (2015).
32. Zhao, K., Ying, Z. & He, S. EMF exposure study concerning mmWave phased array in mobile devices for 5G communication. *IEEE Antennas Wirel. Propag. Lett.* **15**, 1132–1135 (2015).
33. Belrhiti, L., Riouch, F., Tribak, A., Terhzaz, J. & Sanchez, A. M. Investigation of dosimetry in four human head models for planar monopole antenna with a coupling feed for LTE/WWAN/ WLAN internal mobile phone. *J. Microwaves Optoelectron. Electromagn. Appl.* **14**, 1247–1249 (2017).
34. Colombi, D., Thors, B. & Tornevik, C. Implications of EMF exposure limits on output power levels for 5G devices above 6 GHz. *IEEE Antennas Wirel. Propag. Lett.* **14**, 1247–1249 (2015).
35. Laghari, M. R., Hussain, I., Ali Memon, K. & Yaseen, K.G. Modeling and analysis of 5G antenna radiation effect on human head by calculating specific absorption rate (SAR) using adult brain model. *J. Inf. Commun. Technol. Robot. Appl.* **9**, 13–18 (2018).
36. Zada, M., Ali Shah, I. & Yoo, H. Integration of sub-6-GHz and mm-wave bands with a large frequency ratio for future 5G MIMO application. *IEEE Access* **9**, 11241–11251 (2021).
37. Sethi, W. T., Ashraf, M. A., Ragheb, A., Alasaad, A. & Alshebeili, S. A., Demonstration of millimeter wave 5G setup employing high-gain Vivaldi array. *Int. J. Antennas Propag.* **2018**, 3927153, pp. 12. <https://doi.org/10.1155/2018/3927153> (2018).
38. OjaroudiParchin, N. *et al.* MM-wave phased array Quasi-Yagi antenna for the upcoming 5G cellular communications. *Appl. Sci.* **9**(5), 9 (2019).
39. Bang, J. & Choi, J. A compact hemispherical beam-coverage phased array antenna unit for 5G mm-Wave applications. *IEEE Access* **8**, 139715–139726 (2020).
40. Kim, K. W. & Rahmat-Samii, Y. Handset antennas and humans at Ka-band: The importance of directional antennas. *IEEE Trans. Antennas Propag.* **46**(6), 949–950 (1998).

Author contributions

All of the authors wrote the main manuscript text and All authors reviewed the manuscript. A.L. is submitting and Z.A. is corresponding author.

Competing interests

The authors declare no competing interests.

Additional information

Correspondence and requests for materials should be addressed to Z.A.

Reprints and permissions information is available at www.nature.com/reprints.

Publisher's note Springer Nature remains neutral with regard to jurisdictional claims in published maps and institutional affiliations.



Open Access This article is licensed under a Creative Commons Attribution 4.0 International License, which permits use, sharing, adaptation, distribution and reproduction in any medium or format, as long as you give appropriate credit to the original author(s) and the source, provide a link to the Creative Commons licence, and indicate if changes were made. The images or other third party material in this article are included in the article's Creative Commons licence, unless indicated otherwise in a credit line to the material. If material is not included in the article's Creative Commons licence and your intended use is not permitted by statutory regulation or exceeds the permitted use, you will need to obtain permission directly from the copyright holder. To view a copy of this licence, visit <http://creativecommons.org/licenses/by/4.0/>.

© The Author(s) 2021