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# **Distributed coordinated OPEN attitude tracking control of a multi‑spacecraft system with dynamic leader under communication delays**

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**This paper is dedicated to the challenging issue of the cooperative attitude tracking control of a multi-spacecraft system under communication delays. A state estimator using the attitude information of the neighbors is designed for each follower spacecraft to estimate the time-varying attitude information of the leader spacecraft in the case that the leader spacecraft cannot directly communicate with all following spacecraft. By constructing auxiliary variables based on estimated values and proposing a fxed-time control law to ensure that the auxiliary variables of the spacecraft can reach zero in fxed time, which is independent of the initial state, the efects of time-varying reference attitudes on the system can be reduced. The attitude error and the estimation error are proven to converge to the region containing the origin by input-to-state stability theory combined with the Lyapunov–Krasovskii approach. To further illustrate the efectiveness of the proposed control algorithm, numerical simulation results are presented.**

Multi-spacecraft distributed cooperative attitude control, an important branch of cooperative control, has received extensive attention in recent years. Compared with traditional centralized coordination methods, distributed cooperative attitude control has the advantages of high efficiency, strong robustness, and high reliability. Distributed attitude cooperative control is dedicated to synchronizing the attitude of each spacecraf to a common value. In Zhang and Demetriou<sup>1</sup> and Ma et al.<sup>[2](#page-8-1)</sup>, the leaderless cooperative control problem for multiple spacecraft systems was investigated. With this algorithm, the attitude of each spacecraf was synchronized; however, the attitude value of this synchronization was not set in advance, which means that the fnal spacecraf attitude was uncertain. Under the assumption that all followers obtained the leader's acceleration, the coordinated tracking problem with a dynamic leader was investigated in Hong et al.<sup>3</sup>, and the proposed algorithm ensures that the attitude of each spacecraft can track the desired trajectory asymptotically. Du et al.<sup>4</sup> studied the multi-spacecraft attitude tracking problem, and the proposed control algorithms could guarantee that each spacecraf tracks the desired time-varying attitude. Moreover, Zou et al.<sup>[5](#page-8-4)</sup> proposed a cooperative attitude control algorithm without velocity measurement.

The aforementioned literature on leader-follower multi-spacecraft guarantees only the asymptotic convergence of the spacecraf attitude error under the premise that all followers can obtain information from the leader spacecraft. Compared to asymptotic convergence, Liang et al.<sup>6</sup> proposed a finite-time control scheme, which achieved finite-time convergence of spacecraft attitude error under arbitrary communication topologies. To improve the robustness of the system, Zou and Kumar<sup>7</sup> studied the multi-spacecraft distributed attitude coordination control problem in the consideration of external disturbances, and the proposed control algorithms could guarantee that the system achieves finite-time stability. The existing literature on finite-time cooperative attitude control algorithms for multiple spacecraft systems has achieved fruitful results; however, the settling time of finite-time control algorithms is dependent on the initial conditions of the system. In Polyakov<sup>[8](#page-8-7)</sup>, fixed-time convergence was proposed, which guarantees that the settling time is independent of the initial states; thus, the settling time of fixed-time algorithms is dependent on only the design parameters. Shi and Zhou<sup>[9](#page-8-8)</sup> studied the fixed-time

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attitude tracking problem of multi-spacecraf, and the algorithms could guarantee the convergence time of the system states within a fixed time. Following the idea exposed by Ton and Petersen<sup>10</sup>, a disturbance observer was employed in the fxed-time control algorithms to reduce the interference of the multi-spacecraf system. Similarly, taking the external interference and uncertainties of the system into account, the proposed algorithm guaranteed that the system states converge to the regions containing the origin in fxed time in Gao and Ca[i11](#page-8-10).

In addition to the convergence rate of the system, the communication delay, which afects the control performance and even undermines the stability of the system, must be addressed for attitude tracking control. Researchers have made great eforts to study the communication delay problem in linear and nonlinear multia-gent systems. Abdessameud and Tayebi<sup>[12](#page-8-11)</sup> proposed an attitude synchronization scheme that solves the leaderless problem under directed interconnection between rigid bodies with constant communication delays. Similarly, the frequency domain approach was employed to investigate the stability of linear systems under communication delays in Wang<sup>[13](#page-8-12)</sup>. The Lyapunov–Krasovskii function method is an effective approach for nonlinear systems with time-varying delays. In the case that the reference attitude is time-varying and can be obtained by each spacecraf, the back-stepping technique and the Lyapunov–Krasovskii method contribute to the control law design in Guo et al.[14](#page-8-13). Zhou et al.[15](#page-8-14) proposed state feedback attitude synchronization tracking control algorithms and analyzed the delay-dependent stability of the delay system. However, the existing literature on attitude cooperative control for multiple spacecraf with time-varying communication delays focuses on leaderless synchronization or tracking with a dynamic leader under the assumption that the information of the leader is obtained by each follower.

Motivated by the above facts, a distributed attitude cooperative tracking control algorithm is proposed for a multi-spacecraft system with time-varying communication delays. Modeled by the Euler-Lagrange equation, the control objective is to ensure that each follower spacecraft attitude error converges to the region containing the origin when only a part of the follower spacecraft can acquire the dynamic attitude of the leader spacecraft. For the auxiliary variables constructed in this paper, a fxed-time control algorithm is added to make the convergence time of the auxiliary variables independent of the initial state, thereby reducing the infuence of the time-varying reference state on the system.

Compared with existing works, the main contributions of this study are as follows. First, compared with Zhou et al.<sup>15</sup> and Liang et al.<sup>6</sup>, not all follower spacecraft have access to the time-varying attitude information of the leader spacecraft under the directed communication topology. A distributed estimator is employed to estimate the attitude trajectory of the leader for each spacecraf using its neighbors' information under communication delays. Second, compared with Zou and Kumar<sup>[7](#page-8-6)</sup> and Zhao and Jia<sup>16</sup>, in the controller design, a fixed-time control algorithm is combined so that the settling time of the auxiliary variable is independent of the initial state. Tis can reduce the impact of the time-varying reference target state on the overall system. Tird, in contrast to Zhai and Xia<sup>[17](#page-9-0)</sup> in which the algorithm focused on the static leader case, the attitude tracking control algorithms with communication delays in this paper focus on tracking with a dynamic leader spacecraf.

**Notations:**⊗ represents the Kronecker product. For a vector  $x = [x_1, \ldots, x_n]^T$ , we use  $||x||$  to denote the Euclidean norm. Define  $x^{\alpha} = [x_1^{\alpha}, \dots, x_n^{\alpha}]^T$ ,  $[|x_1|^{\alpha}, \dots, |x_n|^{\alpha}]^T$ ,  $sig^{\alpha}(x) = [sgn(x_1)|x_1|^{\alpha}, \dots, sgn(x_n)|x_n|^{\alpha}]^T$  and for  $x = [x_1, x_2, x_3]^T, x^{\times} =$  $\begin{bmatrix} 0 & -x_3 & x_2 \end{bmatrix}$  $x_3$  0 − $x_1$ 1 .

### **Background**

**Euler–Lagrange system.** Consider a multiple rigid spacecraf system consisting of *n* followers, labeled as spacecraft  $i, i = 1, \ldots, n$ , and one leader labeled as spacecraft 0. The attitude of the ith follower is described by modifed Rodriguez parameters (MRPs) as

<span id="page-1-0"></span>
$$
\dot{\sigma}_i = G(\sigma_i)\omega_i,\tag{1}
$$

where  $\sigma_i=[\sigma_{i1},\sigma_{i2},\sigma_{i3}]\in R^3$  is the MRP of the *i*th spacecraft denoting the attitude orientation of the body-fixed frame with respect to the inertial frame.  $\omega_i \in R^3$  is the angular velocity of the *i*th rigid body with respect to the inertial frame expressed in the body frame of the *i*th rigid body.  $G(\sigma_i)$  is described as

$$
G(\sigma_i) = \frac{1}{2} \left[ \frac{1 - \|\sigma_i\|^2}{2} I_3 + \sigma_i^\times + \sigma_i \sigma_i^\mathsf{T} \right]. \tag{2}
$$

The matrix  $G(\sigma_i)$  is provided with following properties

$$
\det(G(\sigma_i)) = \frac{(1 + \sigma_i^2)^3}{4} \neq 0
$$
  
\n
$$
G^T(\sigma_i)G(\sigma_i) = \left(\frac{1 + \sigma_i^2}{4}\right)^2 I_3.
$$
\n(3)

The dynamics of the *i*th spacecraft in the system are described by

 $-x_2$   $x_1$  0

<span id="page-1-3"></span><span id="page-1-2"></span><span id="page-1-1"></span>
$$
J_i \dot{\omega}_i = -\omega_i^{\times} J_i \omega_i + u_i, \tag{4}
$$

where  $J \in R^{3 \times 3}$  is the inertia matrix of the *i*th spacecraft and  $u_i \in R^3$  is the control torque of the *i*th spacecraft. The  $n$  followers are represented with Euler–Lagrange equations by combining  $(1)$  and  $(4)$  $(4)$  as

$$
M_i(\sigma_i)\ddot{\sigma}_i + C_i(\sigma_i, \dot{\sigma}_i)\dot{\sigma}_i = \tau_i, \tag{5}
$$

where  $C_i(\sigma_i, \dot{\sigma}_i) = -G^{-T}(\sigma_i)J_iG^{-1}(\sigma_i)\dot{G}(\sigma_i)G^{-1}(\sigma_i) - G^{-T}(\sigma_i)(J_i\omega_i)^{\times}G^{-1}(\sigma_i)$ ,  $\tau_i = G^{-T}(\sigma_i)u_i$ , and  $M_i(\sigma_i) = G^{-T}(\sigma_i)I_iG^{-1}(\sigma_i)$ . According to ([3\)](#page-1-2),  $M_i(\sigma_i)$  is the symmetric positive-definite inertia matrix. For more details on the Euler-Lagrangian model, please refer to Meng and Ren<sup>18</sup>. Throughout the subsequent analysis, the following fundamental properties of system  $(5)$  $(5)$  $(5)$  are given by Zhai and Xia<sup>17</sup>:

**Property 1** Parameter Boundedness: For any *i*th spacecraft, there exist positive constants  $k_m$ ,  $k_M$  and  $k_C$ ,  $0 < k_m I_3 \leq M_i(\sigma_i) \leq k_M I_3, \|C_i(\sigma_i, \dot{\sigma}_i)\| \leq k_C \|\dot{\sigma}_i\|.$ 

**Property 2** Linearity in the dynamic parameters:  $M_i(\sigma_i)x + C_i(\sigma_i, \dot{\sigma}_i)y = Y_i(\sigma_i, \dot{\sigma}_i, x, y) \Theta_i$ , for any  $x, y \in \mathbb{R}^3$ , where  $\Theta_i$  is the constant parameter vector associated with the *i*th spacecraft and  $Y_i(\sigma_i, \dot{\sigma}_i, x, y)\Theta_i$  is a known regression matrix.

**Graph theory.** Graph theory is briefy introduced to represent the topology of the information fow among multi-spacecraft system. A directed weighted graph can be denoted as  $G = (N, E, A)$ , where  $N = (n_0, n_1, \ldots, n_n)$ is a finite nonempty set of nodes and  $E \subseteq N \times N$  is a set of unordered pairs of nodes. An edge  $(n_i, n_j) \in E$ denotes that the *j*th spacecraft can obtain the information from the *i*th spacecraft and vice versa. The adjacency matrix  $A = [a_{ij}] \in R^{n \times n}$  of graph *G* is defined such that adjacency elements  $a_{ij}$  satisfy if  $(n_i, n_j) \in E$ , and  $a_{ij} = 0$ otherwise. The Laplacian matrix  $L = [l_{ij}] \in R^{n \times n}$  associated with *A* is defined as  $l_{ii} = \sum_{i \neq j} a_{ij}$  and  $l_{ij} = -a_{ij}$ , where  $i \neq j$ . The directed graph *G* has a directed spanning tree if and only if there exists at least one node having a directed path to all other nodes. The leader adjacency matrix is defined as  $B=diag(a_{10}, a_{20}, \ldots, a_{n0})$ , where  $a_{i0}=1$  if the *i*th spacecraft can communicate with the leader spacecraft and  $a_{i0}=0$  otherwise. Define  $H = L + B$ .

<span id="page-2-5"></span>**Assumption 1** The communication topology graph  $\bar{G}$  of the multi-spacecraft system has a directed spanning tree.

<span id="page-2-4"></span>**Lemma 2.1** (Ren and Cao<sup>[19](#page-9-2)</sup>) If graph G has a directed spanning tree, all eigenvalues of matrix H have positive real *parts*.

**Mathematic background.** Consider the following system:

<span id="page-2-0"></span>
$$
\dot{x} = g(t, x), x(0) = x_0,\tag{6}
$$

where  $x \in R^n$  and  $g: R_+ \times R^n \to R^n$  is a nonlinear function, which can be discontinuous. For system ([6](#page-2-0)), the following defnitions and lemmas are given.

Definition 1 (*Polyakov*<sup>[8](#page-8-7)</sup>) The equilibrium of system [\(6\)](#page-2-0) is fixed-time stable if it is finite-time stable, and the settling time  $T(x_0)$  is uniformly bounded for any initial states, that is,  $\exists T_{\text{max}} > 0$ , such that  $T(x_0) \leq T_{\text{max}}$ ,  $\forall x_0 \in \mathbb{R}^n$ .

<span id="page-2-3"></span>**Lemma 2.2** (Polyakov<sup>8</sup>) If there exists a continuous radially unbounded function  $V: R^n \to R_+ \cup \{0\}$  such that

- (1)  $V(x) = 0 \Leftrightarrow x = 0;$
- (2) *for any solution x(t)* of [\(6\)](#page-2-0) *satisfies the inequality*  $D^*V(x(t)) \leq -(\alpha V^p(x(t)) + \beta V^q(x(t)))^k$  *for some*  $\alpha$ ,  $\beta$ ,  $p$ ,  $q$ ,  $k > 0$ :  $pk < 1$ ,  $qk > 1$ , then the origin is globally fixed-time stable for system ([6](#page-2-0)) and the following *estimate holds*:

$$
T(x_0) \le \frac{1}{\alpha^k (1 - pk)} + \frac{1}{\beta^k (qk - 1)} \forall x_0 \in R^n,
$$
\n(7)

*where*  $D^* f(t) = \lim_{h \to 0^+} \sup \frac{f(t+h)-f(t)}{h}$ .

<span id="page-2-7"></span>**Lemma 2.3** (Khalil<sup>[20](#page-9-3)</sup>) *Consider the system* 

<span id="page-2-1"></span>
$$
\dot{x} = f(t, x, u),\tag{8}
$$

*where f*(*t*, *x*, *u*) *is continuously diferentiable and globally Lipschitz in* (*x*, *u*), *uniformly in t. If the unforced system*   $\dot{x} = f(t, x, 0)$  has a globally exponentially stable equilibrium point at the origin  $x = 0$ , then the system ([8\)](#page-2-1) is input*to-state stable*.

<span id="page-2-2"></span>**Lemma 2.4** (Zuo and Lin<sup>21</sup>) Let  $\xi_1, \xi_2, \ldots, \xi_n \geq 0$ . Then

$$
\sum_{i=1}^{n} \xi_i^p \ge \left(\sum_{i=1}^{n} \xi_i\right)^p, \text{ if } 0 < p \le 1,
$$
\n
$$
\sum_{i=1}^{n} \xi_i^p \ge n^{1-p} \left(\sum_{i=1}^{n} \xi_i\right)^p, \text{ if } p > 1.
$$
\n
$$
(9)
$$

<span id="page-2-6"></span>**Lemma 2.5** (Wang et al.<sup>[22](#page-9-5)</sup>) *For any scalar*  $\gamma_2 > \gamma_1 > 0$ *, symmetric positive definite matrix P*  $\in R^{m \times m}$  *and vector function*  $\varpi : [\gamma_1, \gamma_2] \to \mathbb{R}^m$  such that the following integration holds:

$$
\int_{\gamma_1}^{\gamma_2} \varpi^T(\sigma) P \varpi(\sigma) d\sigma \ge \frac{1}{\gamma_1 \sigma} \left[ \int_{\gamma_1}^{\gamma_2} \varpi(\sigma) d\sigma \right]^T P \left[ \int_{\gamma_1}^{\gamma_2} \varpi(\sigma) d\sigma \right],\tag{10}
$$

*where*  $\gamma_{12} = \gamma_2 - \gamma_1$ .

<span id="page-3-2"></span>**Lemma 2.6** (Gahinet and Apkarian<sup>23</sup>) The block matrix 
$$
\begin{bmatrix} A_1 & B \\ B^T & A_2 \end{bmatrix} \le 0
$$
 if and only if  $\begin{cases} A_1 \le 0 \\ A_2 - B^T A_1^{-1} B \le 0 \end{cases}$ 

# **Coordinated attitude tracking control with a dynamic leader under communication delays**

In this section, we address the distributed coordinated attitude tracking control problem for multi-spacecraf under directed communication graph with a dynamic leader under communication delays. The attitude of the leader spacecraft is generated as $^{24}$  $^{24}$  $^{24}$ 

 $\sigma$ 

<span id="page-3-0"></span>
$$
\dot{v} = Qv
$$
  
(11)  

$$
v_0 = Nv,
$$

where  $O \in R^{m \times m}$  and  $N \in R^{3 \times m}$  are constant matrices and  $v \in R^m$  is an auxiliary variable.

*Remark 1* System [\(11](#page-3-0)) can generate multiple types of  $\sigma_0$ , such as step signals of arbitrary amplitude, ramp signals of arbitrary slope, and sinusoidal signals of arbitrary amplitudes and initial phases. By the reasonable design of *N*, a linear combination of *v* will generate a variety of different signals<sup>25</sup>. We assume  $m = 3$  in this paper.

Defne the attitude tracking error and its derivative of each follower spacecraf as

<span id="page-3-1"></span>
$$
\tilde{\sigma}_i = \sigma_i - \sigma_0 \n\dot{\tilde{\sigma}}_i = \dot{\sigma}_i - \dot{\sigma}_0,
$$
\n(12)

where  $i = 1, 2, ..., n$ , and the control objective of this paper is to ensure  $\lim_{t \to \infty} \tilde{\sigma}_i(t) = 0$ ,  $\lim_{t \to \infty} \tilde{\sigma}_i(t) = 0$ . Because of the existence of time delay in communication between spacecraft, the distributed estimator is proposed as

$$
\dot{\hat{v}}_i(t) = Q\hat{v}_i(t) - \alpha \sum_{j=0}^n a_{ij} [\hat{v}_i(t-T) - \hat{v}_j(t-T)],
$$
\n(13)

where  $\hat{v}_i(t)$  is the estimate of attitude information  $v_0(t)$ ,  $\alpha$  is a positive constant and. *T* is the time-varying communication delay.

*Remark 2* In this paper, the communication time delay between the available spacecraft is considered. There is  $\hat{v}_i(t - T)$  in [\(13](#page-3-1)), however, this item is not caused by the communication time delay, but the states of the *i*th spacecraft at the time  $t - T$  is called during the controller calculation process.

<span id="page-3-3"></span>**Assumption 2** The time-varying communication delay *T* satisfies  $0 \le T \le T_0$  and  $0 \le T(t) \le d < 1$ , where  $T_0$  is a positive constant and there exist positive definite matrices  $P_1$ ,  $P_2$ , *W* and reasonably matrices *Z* such that

$$
A=\begin{bmatrix}P_1+2(Q\otimes I_n)+dW+d(Q\otimes I_n)^TP_2(Q\otimes I_n)+2Z-\alpha[H\otimes I_3+d(Q\otimes I_n)^TP_2(H\otimes I_3)]-Z\\-\alpha[H\otimes I_3+d(Q\otimes I_n)^TP_2(H\otimes I_3)]-Z]^T&-(1-d)P_1+d\alpha^2(H\otimes I_3)^TP_2(H\otimes I_3)\end{bmatrix}\leq 0.
$$

**Remark 3** According to Lemma [2.6](#page-3-2), let  $A = \begin{bmatrix} A_{11} & A_{12} \\ A_{12}^T & A_{22} \end{bmatrix}$ , it is feasible to make matrix  $A_{11}$  a negative definite

matrix by choosing matrix *Z* reasonably. The positive definite  $P_1$ ,  $P_2$  and *H* could make matrix  $A_{22}$  a negative definite matrix; thus,  $A_{22} - A_{12}T A_{11}^{-1}A_{12} \le 0$  is feasible. For example, we can choose a negative definite matrix *Z* reasonably. Compared with Wang et al.<sup>[26](#page-9-9)</sup>, although the LMI forms are similar, the design of this paper can guarantee each block matrix on main diagonal negative defnite, thus guaranteeing the correctness of the proof. We first give the boundedness proof for the relevant states [\(13\)](#page-3-1). Consider the following Lyapunov–Krasovskii functional

$$
V_1 = \hat{v}^T(t)\hat{v}(t) + \int_{t-T}^t \hat{v}^T(\zeta)P_1\hat{v}(\zeta)d\zeta + \int_{-T}^0 \int_{t+\theta}^t \hat{v}^T(\zeta)P_2\hat{v}(\zeta)d\zeta d\theta + \int_{0}^t \int_{\theta-T}^{\theta} \left[ v^T(\theta) \hat{v}^T(\zeta) \right] \left[ \begin{array}{c} W & Z^T \\ Z & P_2 \end{array} \right] \left[ \begin{array}{c} \hat{v}(\theta) \\ \hat{v}(\zeta) \end{array} \right] d\zeta d\theta,
$$
\n(14)

where  $\hat{v}(t)$  is the column vector of column  $\hat{v}_i(t)$ , and the derivative of  $V_1$  is

<span id="page-3-4"></span>4

$$
\dot{V}_{1} = 2\hat{v}^{T}(t)\dot{\hat{v}}(t) + \hat{v}(t)P_{1}\hat{v}(t) - (1 - T)\hat{v}^{T}(t - T)P_{1}\hat{v}(t - T) - \int_{t-T}^{t} \dot{\hat{v}}^{T}(\zeta)P_{2}\dot{\hat{v}}(\zeta)d\zeta \n+ T\dot{\hat{v}}(t)P_{2}\dot{\hat{v}}(t) + T\hat{v}^{T}(t)W\hat{v}(t) + 2\hat{v}^{T}(t)Z[\hat{v}(t) - \hat{v}(t - T)] + \int_{t-T}^{t} \dot{\hat{v}}^{T}(\zeta)P_{2}\dot{\hat{v}}(\zeta)d\zeta \n\leq \hat{v}^{T}(t)[-2\alpha(H \otimes I_{3}) - 2\alpha d(Q \otimes I_{n})^{T}P_{2}(H \otimes I_{3}) - 2Z]\hat{v}(t - T) + \hat{v}(t)[P_{1} + 2(Q \otimes I_{n}) \n+ dW + d(Q \otimes I_{n})^{T}P_{2}(Q \otimes I_{n}) + 2Z]\hat{v}(t) - \hat{v}^{T}(t - T)[(1 - d)P_{1} - d\alpha^{2}(H \otimes I_{3})^{T}P_{2}(H \otimes I_{3})]\hat{v}(t - T) \n= [\hat{v}^{T}(t)\hat{v}^{T}(t - T)]A[\hat{v}^{T}(t)\hat{v}^{T}(t - T)]^{T} \leq 0
$$
\n(15)

therefore,  $\hat{v}(t)$  is bounded according to Assumption [2](#page-3-3) and ([14](#page-3-4)).

Defne auxiliary variables as

$$
\dot{\sigma}_{ri}(t) = NQ\hat{v}_i(t) - \beta \sum_{j=0}^{n} a_{ij} [\sigma_i(t-T) - \sigma_j(t-T)] \tag{16}
$$

<span id="page-4-3"></span><span id="page-4-2"></span><span id="page-4-1"></span><span id="page-4-0"></span>
$$
s_i(t) = \dot{\sigma}_i(t) - \dot{\sigma}_{ri}(t),
$$
\n(17)

where  $\beta$  is a positive constant. We obtain by Property 2,

$$
M_i(\sigma_i)\ddot{\sigma}_{ri} + C_i(\sigma_i, \dot{\sigma}_i)\dot{\sigma}_{ri} = Y_i(\sigma_i, \dot{\sigma}_i, \ddot{\sigma}_{ri}, \dot{\sigma}_{ri})\Theta_i.
$$
\n(18)

Combining [\(5\)](#page-1-3) and [\(18](#page-4-0)), the coordinated attitude tracking control law for the *i*th follower spacecraf is proposed as

$$
\tau_i = C_i(\sigma_i, \dot{\sigma}_i)s_i + Y_i \Theta_i - k_1 M_i(\sigma_i) sig^p(s_i) - k_2 M_i(\sigma_i) sig^q(s_i) - k_3 s_i,
$$
\n(19)

where  $k_1$  and  $k_2$  are positive constants,  $0 < p < 1$  and  $q > 1$ .

<span id="page-4-4"></span>**Theorem 1** *Considering the multi-spacecraf attitude system* [\(5\)](#page-1-3), *the auxiliary variable* si *of each follower spacecraf converges to zero in fxed time by properly designing the parameter coordinated control law* ([19](#page-4-1)).

*Proof* We have shown that the states  $\hat{v}$  are bounded in ([14](#page-3-4)); moreover,  $\dot{\hat{v}}_i$ ,  $\dot{\sigma}_{ri}$  and  $\ddot{\sigma}_{ri}$  are bounded from ([13](#page-3-1)) and [\(16\)](#page-4-2). By Property 1,  $Y_i(\sigma_i, \dot{\sigma}_i, \ddot{\sigma}_{ri}, \dot{\sigma}_{ri})\Theta_i$  is bounded for bounded states  $\ddot{\sigma}_{ri}$  and  $\dot{\sigma}_{ri}$ .  $\tau$ , sand  $\Theta$  are column stack vectors of  $\tau_i$ ,  $s_i$  and  $\Theta_i$  respectively.  $C(\sigma, \dot{\sigma})$ ,  $M(\sigma)$  and *Y* are block diagonal matrices of  $C_i(\sigma_i, \dot{\sigma}_i)$ ,  $M_i(\sigma_i)$  and  $Y_i$ , respectively. Substituting [\(17\)](#page-4-3), [\(18\)](#page-4-0) and ([19\)](#page-4-1) into [\(5\)](#page-1-3), the system can be expressed as

$$
M(\sigma)\dot{s} = -k_1 M(\sigma) sig^p(s) - k_2 M(\sigma) sig^q(s) - k_3 s. \tag{20}
$$

Consider the Lyapunov function candidate as

$$
V_2 = \frac{1}{2}s^T s. \tag{21}
$$

Taking the time derivative of the Lyapunov function  $V_2$ , by means of Lemma [2.4,](#page-2-2) we obtain

$$
\dot{V}_2 = s^T \dot{s} \n= -s^T M^{-1}(\sigma)(k_1 M(\sigma) sig^p(s) + k_2 M(\sigma) sig^q(s) + k_3 s) \n= -k_1 s^T sig^p(s) - k_2 s^T sig^q(s) - k_3 s^T M^{-1}(\sigma) s \n\leq -k_1 \sum_{i=1}^n \sum_{l=1}^3 (s_{il}^2)^{\frac{p+1}{2}} - k_2 \sum_{i=1}^n \sum_{l=1}^3 (s_{il}^2)^{\frac{q+1}{2}} \n\leq -k_1 [\sum_{i=1}^n \sum_{l=1}^3 (s_{il}^2)]^{\frac{p+1}{2}} - k_2 (3n)^{\frac{1-q}{2}} [\sum_{i=1}^n \sum_{l=1}^3 (s_{il}^2)]^{\frac{q+1}{2}} \n\leq -k_1 \cdot 2^{\frac{p+1}{2}} V_2^{\frac{p+1}{2}} - (3n)^{\frac{1-q}{2}} k_2 \cdot 2^{\frac{q+1}{2}} V_2^{\frac{q+1}{2}}.
$$
\n(22)

Since  $M(\sigma)$  is positive definite, we can obtain  $V_2 < 0$ , which means that  $s_i$  is bounded. According to the Property 1 and Lemma [2.2](#page-2-3), the auxiliary variable  $s_i$  for the *i*th spacecraft will converge to zero in fixed time with a settling time

<span id="page-4-5"></span>
$$
T_s \le \frac{2^{\frac{1-p}{2}}}{k_1(1-p)} + \frac{\left(\frac{2}{3n}\right)^{\frac{1-q}{2}}}{k_2(q-1)}.
$$
\n(23)

The proof of Theorem [1](#page-4-4) is complete.  $\Box$ 

Define  $\tilde{v}_i(t) = \hat{v}_i(t) - v(t)$ .  $\tilde{\sigma}(t)$  and  $\tilde{v}(t)$  are the column stack vectors of  $\tilde{\sigma}_i(t)$  and  $\tilde{v}_i(t)$ , respectively. Rewrite [\(13\)](#page-3-1) and ([16](#page-4-2)) as

$$
\dot{\tilde{v}}(t) = (Q \otimes I_n)\tilde{v}(t) - \alpha(H \otimes I_3)\tilde{v}(t - T),
$$
\n(24)

$$
\dot{\tilde{\sigma}}(t) = (NQ \otimes I_n)\tilde{\nu}(t) - \beta(H \otimes I_3)\tilde{\sigma}(t - T) + s. \tag{25}
$$

Consider the following system

$$
\begin{cases} \dot{\tilde{\sigma}}(t) = (NQ \otimes I_n)\tilde{v}(t) - \beta(H \otimes I_3)\tilde{\sigma}(t - T) \\ \dot{\tilde{v}}(t) = (Q \otimes I_n)\tilde{v}(t) - \alpha(H \otimes I_3)\tilde{v}(t - T) \end{cases}
$$
(26)

define 
$$
R = \begin{bmatrix} \beta(H \otimes I_3) & 0 \\ 0 & \alpha(H \otimes I_3) \end{bmatrix}
$$
,  $E = \begin{bmatrix} 0 & NQ \otimes I_n \\ 0 & Q \otimes I_n \end{bmatrix}$ ,  $F = \alpha(H \otimes I_3) - Q \otimes I_n$  and  $U = R - E$ . Let

 $x = \left[\tilde{\sigma}^{\,T}, \tilde{v}^T\right]^{\,T}$ . The system can be written in the following form

<span id="page-5-4"></span><span id="page-5-1"></span><span id="page-5-0"></span>
$$
\dot{x}(t) = Ex(t) - Rx(t - T).
$$
 (27)

*Remark 4* According to Lemma [2.1](#page-2-4), all eigenvalues of matrix *H* have positive real parts. In the case of the estab-lishment of Assumption [1](#page-2-5), matrix *F* is positive definite by reasonably selecting positive constant  $\alpha$ . Thus, the block matrix *U* can be a positive defnite matrix.

<span id="page-5-5"></span>**Theorem 2** *Consider the linear system with communication delays described by* [\(27\)](#page-5-0). *If there exist positive defnite matrices* P3, P4 *and* P5 *such that the following linear matrix inequality (LMI) holds*:

$$
B = \begin{bmatrix} -2P_3U + dP_4 & P_3R + (1-d)P_4 \\ R^T P_2 + (1-d)P_4 & -P_5 - (1-d)P_4 \end{bmatrix} + T_0^2 \begin{bmatrix} -U^T \\ R^T \end{bmatrix} R^T \begin{bmatrix} -U & R \end{bmatrix} < 0 \tag{28}
$$

*then system state x converges to zero asymptotically*. *Under Assumptions* [1](#page-2-5) *and* [2](#page-3-3), *the distributed estimator* [\(13\)](#page-3-1) *and the coordinated control law* [\(19\)](#page-4-1), *which depends on* [\(16](#page-4-2)) *and* ([17\)](#page-4-3), *can guarantee that the attitude tracking error of the follower spacecraft asymptotically converges to zero.* 

*Remark 5* The form of LMI [\(28](#page-5-1)) is inspired by (Polyak et al.<sup>27</sup>). Similar to the analysis in Remark 3, the corresponding solutions  $P_3$ ,  $P_4$  and  $P_5$  can be obtained by the mature solution method of linear matrix inequality (Last<sup>28</sup>). In LMI ([28\)](#page-9-11), the upper bound  $T_0$  of the communication delay that the system can cope with only related to design parameters and not to the structure of the controller itself. Therefore,  $T_0$  does not affect the proof of system stability. But in practice, the size of  $T_0$  affects the selection of  $\alpha$  and  $\beta$ .

*Proof* Consider the following Lyapunov–Krasovskii functional for system [\(27\)](#page-5-0):

$$
V_3 = x^T(t)P_3x(t) + \int_{t-T}^t x^T(\zeta)P_4x(\zeta)d\zeta + T \int_{-T}^0 \int_{t+\theta}^t \dot{x}^T(\zeta)P_5\dot{x}(\zeta)d\zeta d\theta.
$$
 (29)

Taking the time derivative of the Lyapunov–Krasovskii function  $V_3$ , we have

$$
\dot{V}_3 = 2x^T(t)P_3\dot{x}(t) + x^T(t)P_4x(t) - (1 - \dot{T})x^T(t - T)P_4x(t - T) + T^2\dot{x}^T(t)P_5\dot{x}(t) \n- T\int_{t-T}^t \dot{x}^T(\zeta)P_5\dot{x}(\zeta)d\zeta.
$$
\n(30)

By employing the Leibniz–Newton formula, we obtain  $\int_{t-T}^{t} \dot{x}(\zeta) d\zeta = x(t) - x(t-T)$ . Define  $\bar{x}(t) = x(t) - x(t-T)$ . It can be obtained that the following inequality holds according to Lemma [2.5](#page-2-6)

$$
\bar{x}^{T}(t)P_{5}\bar{x}(t) \leq T\int_{t-T}^{t} \dot{x}^{T}(\zeta)P_{5}\dot{x}(\zeta)d\zeta.
$$
\n(31)

Under Assumption [2](#page-3-3), the following inequality can be obtained by using ([31\)](#page-5-2):

$$
\dot{V}_{3} \leq 2x(t)^{T}P_{3}[Ex(t) - Rx(t - T)] + x^{T}(t)P_{4}x(t) - (1 - \dot{T})x^{T}(t - T)P_{4}x(t - T) + T_{0}^{2}\dot{x}^{T}(t)P_{5}\dot{x}(t) \n- \bar{x}^{T}(t)P_{5}\bar{x}(t) \n\leq 2x^{T}(t)P_{3}[(E - R)x(t) + R\bar{x}(t)] + x^{T}(t)P_{4}x(t) - (1 - d)(x(t) - \bar{x}(t))^{T}P_{4}(x(t) - \bar{x}(t)) \n+ T_{0}^{2}[(E - R)x(t) + R\bar{x}(t)]^{T}P_{5}[(E - R)x(t) + R\bar{x}(t)] - \bar{x}^{T}(t)P_{5}\bar{x}(t) \n\leq x^{T}(t)[-2P_{3}U + dP_{4}]x(t) + 2x^{T}(t)[P_{3}R + (1 - d)P_{4}]\bar{x}(t) - \bar{x}^{T}(t)[P_{5} + (1 - d)P_{4}]\bar{x}(t) \n+ T_{0}^{2}[-Ux(t) + R\bar{x}(t)]^{T}P_{5}[-Ux(t) + R\bar{x}(t)] \n= \left[x^{T}(t)\bar{x}(t)^{T}\right]B\left[x^{T}(t)\bar{x}(t)^{T}\right]^{T} < 0.
$$

By substituting ([28](#page-5-1)) into [\(32\)](#page-5-3), linear system ([27](#page-5-0)) is asymptotically stable, which means that  $\lim_{\alpha \to 0} \tilde{\sigma}(t) = 0$  and  $\lim_{t\to\infty}\tilde{v}(t)=0$ . Since the linear system [\(27](#page-5-0)) is asymptotically convergent, its characteristic roots all have negative real parts, so it is exponentially convergent. It thus follows that when  $s = 0$ , the system given by [\(24](#page-4-5)) and ([25](#page-5-4)) is

<span id="page-5-3"></span><span id="page-5-2"></span>(32)



<span id="page-6-0"></span>**Figure 1.** Communication topology.



<span id="page-6-1"></span>**Table 1.** Initial parameters of follower spacecraf.

globally exponentially stable at the origin  $[\tilde{\sigma}^T, \tilde{v}^T]^T = 0$ . Combining that *s*(*t*) has been proven to be bounded and converge to zero in fixed time in Theorem [1](#page-4-4), the system given by [\(24\)](#page-4-5) and ([25](#page-5-4)) is input-to-state stable with respect to the input *s*(*t*) from Lemma [2.3.](#page-2-7) From Theorem [1](#page-4-4),  $V_2 \rightarrow 0$  as  $t \rightarrow T_s$ , i.e.,  $s(t) \rightarrow 0$  as  $t \rightarrow T_s$ . Since systems [\(24\)](#page-4-5) and [\(25\)](#page-5-4) are input-to-state stable with respect to the input *s*(*t*) and the state  $\left[\tilde{\sigma}(t)^T, \tilde{v}(t)^T\right]^T, \dot{V}_3 \to 0$ as  $t \to \infty$ , i.e.,  $\tilde{\sigma}(t) \to 0$  and  $\tilde{\nu}(t) \to 0$  as  $t \to \infty$ . This completes the proof of Theorem [2](#page-5-5).

# **Simulation results**

In this section, to demonstrate the validity of the proposed control algorithms, numerical simulations for a multi-spacecraft system are conducted, and the communication topology is described by Fig. [1](#page-6-0). The system consisting of four follower spacecraft and one leader spacecraft satisfies Assumption [1.](#page-2-5) Table [1](#page-6-1) lists the inertia matrix and initial conditions of the follower spacecraft. The time-varying communication delay *T* is set as  $T = 0.1 + 0.1 \sin(t)$ , which satisfies Assumption [2](#page-3-3). The parameters of the distributed estimator ([13](#page-3-1)), the auxiliary variable  $\dot{\sigma}_{ri}$  and the controller ([19](#page-4-1)) are set as  $\alpha=1$ ,  $\beta=1$ ,  $k_1=0.8$ ,  $k_2=1$ ,  $k_3=1$ ,  $p=0.4$  and  $q=2$ , respectively.  $\lceil -0.005 \times \sin(0.1t) \rceil$ 1

The attitude  $\sigma_0$  of the leader is generated by the following parameters: $v(t)$  =  $0.008 \times \cos(0.1t)$  $-0.007 \times \sin(0.1t)$ 

$$
Q = \begin{bmatrix} 0 & -\frac{1}{16} & 0 \\ \frac{1}{5} & 0 & \frac{1}{10} \\ \frac{1}{5} & -\frac{7}{80} & -\frac{1}{7} \end{bmatrix} N = \begin{bmatrix} -2 & 0 & 0 \\ 0 & 1.6 & 0 \\ 0 & 0 & -2 \end{bmatrix}.
$$

 $\tilde{\sigma}_i(t)$ ,  $\tilde{v}_i(t)$  and  $s_i(t)$  are illustrated in Figs. [2](#page-7-0), [3](#page-7-1) and [4](#page-7-2), respectively. Figure [5](#page-8-16) shows the control torques of the follower spacecraft. The attitude tracking errors  $\tilde{\sigma}_i(t)$ , estimation errors  $\tilde{v}_i(t)$  and auxiliary variable  $s_i(t)$  can converge to region  $|\tilde{\sigma}_{iI}| < 7 \times 10^{-4}$ ,  $|s_{iI}| < 6 \times 10^{-5}$  in 8 s and  $|\tilde{v}_{iI}| < 6 \times 10^{-4}$  in 5 s. As the desired attitude is dynamic, the control torques will never be zero, as shown in Fig. [5](#page-8-16). Afer the multi-spacecraf system reaches attitude synchronization, the control torque tends to be stable. Although the fxed time control algorithm cannot consistently converge the error of the auxiliary variable to zero, as shown in Fig. [4](#page-7-2), it can still be used to reduce the convergence time in the case of tracking the dynamic target, as the convergence time is infuenced by the initial conditions in the traditional control process, which continuously change due to the dynamic reference attitude.

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<span id="page-7-0"></span>**Figure 2.** Attitude tracking errors  $\tilde{\sigma}_i(t)$ .



<span id="page-7-1"></span>



<span id="page-7-2"></span>



<span id="page-8-16"></span>**Figure 5.** Control torques  $u_i(t)$ .

### **Conclusions**

This paper studies the problem of distributed coordinated control of spacecraft attitude tracking under timevarying communication delays, where the communication topology is described as a directed graph and the reference attitude is dynamic. Since the reference attitude information is available to only some of the following spacecraf, a distributed estimator that uses neighbor information under communication delay is proposed to estimate the reference attitude information. An auxiliary variable is designed based on the estimated information. By means of the fxed-time control method, a fxed-time control algorithm is proposed that achieves convergence of the auxiliary variable with fxed time, and the convergence time is independent of the initial state. Tis means that the infuence of the constantly changing initial conditions on the system is reduced. By using the Lyapunov–Krasovskii functional approach and input-to-state stability theory, the proposed control algorithms guarantee that the attitude of each following spacecraf can be asymptotically synchronized with the dynamic attitude of the leader spacecraf.

### **Data availability**

All relevant data are in the manuscript and the relevant literature cited in the manuscript.

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# **Author contributions**

Z.Z. wrote the main manuscript. Z.Z. and Y.W. are responsible for guiding and verifying. All authors reviewed the manuscript.

# **Competing interests**

The authors declare no competing interests.

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