

Algorithm for the classification of multi-modulating signals on the electrocardiogram

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Abstract This article discusses the algorithm to measure electrocardiogram (ECG) and respiration simultaneously and to have the diagnostic potentiality for sleep apnoea from ECG recordings. The algorithm is composed by the combination with the three particular scale transform of $a_j(t)$, $u_j(t)$, $o_j(a_j)$ and the statistical Fourier transform (SFT). Time and magnitude scale transforms of $a_j(t)$, $u_j(t)$ change the source into the periodic signal and $\tau_j = o_j(a_j)$ confines its harmonics into a few instantaneous components at τ_j being a common instant on two scales between t and τ_j . As a result, the multi-modulating source is decomposed by the SFT and is reconstructed into ECG, respiration and the other signals by inverse transform. The algorithm is expected to get the partial ventilation and the heart rate variability from scale transforms among $a_j(t)$, $a_{j+1}(t)$ and $u_{j+1}(t)$ joining with each modulation. The algorithm has a high potentiality of the clinical checkup for the diagnosis of sleep apnoea from ECG recordings.

Keywords Fourier transform · Scale transform · Electrocardiogram · Respiratory waveform · Heart rate variability

1 Introduction

The electrocardiogram (ECG) contains well-known PQRST(U) pattern. Its morphology has been extensively studied since the technique of recording was

introduced in the beginning of the twentieth century. The use of computers for ECG recording enabled studies of dynamic properties of the ECG. In particular, two important characteristics have been revealed: amplitude and frequency modulation.

The modulation of the cardiac cycle is extensively studied based on the signal that is generated from intervals between consecutives R-peaks, known as heart rate variability (HRV). Numerous approaches have been proposed and the signal have been decomposed in various ways—analyzing their spectral properties by frequency and time-frequency methods such as Fourier transform [2], autoregressive spectral estimation [11], wavelet transform [9, 17], or their scaling properties [6].

The amplitude characteristics and amplitude modulation received less attention, especially because the notion of the ECG amplitude is ambiguous: the heart as a source of the signal is moving and the signal is collected from the surface of the body that is moving as well. In addition, it is usually recorded as a 2-D projection. However, it has long been recognized that one of the sources of the amplitude modulation—the respiratory related movement of the thorax can be usefully used to extract the information about the respiration.

The influence of respiration on recording cardiac potentials was reported as early as in 1967; Flaherty et al. [5] and Moody et al. [13] were probably the first to discuss the possibility to derive the respiratory signal from the ECG. Since then several algorithms were proposed to reconstruct respiration by demodulating the ECG signal (e.g. [1, 4, 8, 10]).

The ECG-derived respiration signal either based on changes in heart rate or based on direct effect of res-

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piration on the ECG waveform has been particularly used in studies of obstructive sleep apnoea [3, 16].

The problem to reconstruct waveform from the discrete Fourier analysis is that the biological signal has complicated properties; being non-stationary wave, including random noise and having interactive components. However, though people rarely notice, the essential problem is that there is no way to gather impartial and statistical samples. That is why samples from an unique sampling frequency are biased for spectral analysis even if fine sampling.

The purpose of this study is to introduce an algorithm to classify the multi-modulating waveforms on ECG signals. The algorithm is composed by the three particular scale transforms and the statistical Fourier transform (SFT) with impartial and statistical samples [12]. As a result, the multi-modulating source is decomposed by the SFT and is reconstructed into ECG, respiration and the other signals simultaneously by particular transformations.

2 Algorithm

The principle of this algorithm is the selection of three particular scale transforms. Two scales convert modulating waveforms into the periodic signal and the other converts the periodic components into a few terms. The role of the SFT is to hold sample information for any transforms.

2.1 Scale transforms $a_j(t)$, $u_j(t)$ and $o_j(t)$ on modulating signal

Since a source sample at any instant t_k has the mutual phase relationship of instantaneous components on the SFT, any scale transform including permutation expresses the same sample as own instantaneous components by each scale (Appendix 1).

Three scale transforms $a_j(t)$, $u_j(t)$, $o_j(t)$ are adopted to the periodicity and to the reduction of Fourier components.

$a_j = a_j(t)$ is a scale to match time intervals in the standard's intervals and is called adaptive scale and $u_j = u_j(\tau_j)$ is a scale to match the magnitude spans in the standard's magnitudes. Those scales convert the source $x'(t)$ into the standard $x_j(t)$ as $x_j(t) = x'(a_j)/u_j(a_j)$. $a_j(t)$ and $u_j(a_j)$, then, are defined as a join of each $a_{kj}(t)$ and $u_{kj}(a_{kj})$ ($k = 1, \dots, k_j$) when the standard repeats k_j times in a window T_j .

$o_j = o_j(t)$, called inherent scale, is a timescale to convert the standard $x_j(t)$ into the Fourier series of r_j

components passing through $2r_j$ poles of $x_j(t)$. The scale transform $\tau_j(t)$ is defined as the composite transform as follows (Appendix 2):

$$\tau_j(t) = o_j(a_j(t)) \tag{1}$$

The transformed signal $(x'(\tau_j) - \varepsilon_j(\tau_j))/u_j(\tau_j)$ occupies r_j instantaneous components of the skipped ik_j th term in a window T_j including k_j modulations. While, instantaneous components of the $\varepsilon_j(\tau_j)/u_j(\tau_j)$ scatters into broad components from the 1st to the n th terms. As a result, $(x'(\tau_j) - \varepsilon_j(\tau_j))/u_j(\tau_j)$ fits the scales τ_j , u_j and $\varepsilon_j(\tau_j)/u_j(\tau_j)$ unfits scales τ_j , u_j , and then, the source $x'(t)$ is decomposed into two class modulations of $x'(\tau_j) - \varepsilon_j(\tau_j)$ and $\varepsilon_j(\tau_j)$ (Appendix 3).

That is, $x'(\tau_j) - \varepsilon_j(\tau_j)$ is extracted by $v_j(\tau_j) = \sum u_j(\tau_j) h_{ij}(\tau_j)$ from the ik_j th ($i = 1, \dots, r_j$) components and $\varepsilon_j(\tau_j)$ is extracted by the other $\varepsilon_j(\tau_j) = \sum u_j(\tau_j) h_{ij}(\tau_j)$ except the ik_j th components when n instantaneous components of $x'(\tau_j)/u_j(\tau_j)$ are defined as $h_{ij}(\tau_j)$. The source $x'(t)$, therefore, is decomposed into $x'(t) - \varepsilon_j(t)$ and $\varepsilon_j(t)$ by inverse scale transform $t = t(\tau_j)$ under an adequate $u_j(\tau_j)$.

The magnitude scale $u_{kj}(\tau_j)$ is defined as an expression of the magnitude pattern in each window T_{kj} when the signals $v_j(\tau_j)(=x'(\tau_j) - \varepsilon_j(\tau_j))$, $\varepsilon_j(\tau_j)$ and $u_j(\tau_j)$ are expressed by $\{v_{kj}(\tau_j)\}_k$, $\{\varepsilon_{kj}(\tau_j)\}_k$ and $\{u_{kj}(\tau_j)\}_k$ ($k = 1, \dots, k_j$), and each changes its waveform independently. Especially, $v_j(\tau_j)$ is expressed by a constant value $u_j(\tau_j) = u_{kj}$ in each window T_{kj} when its magnitude modulation takes a similar figure with the standard (Appendix 4).

2.2 Iterative scale classification of modulating signals

The timescale transform of the residual $\varepsilon_j(t)$ is defined as $\tau_{j+1}(t) = o_{j+1}(a_{j+1}(t))$ when several waves of $\{\varepsilon_{kj}(t)/u_{kj+1}(t)\}_k$ in $\varepsilon_j(t)/u_{j+1}(t)$ reappears within a window $T_{j+1}(T_{j+1} \leq T_j)$. $\varepsilon_j(t)$ is analyzed as the $j + 1$ th modulation with the iterative algorithm as $\varepsilon_j(t) - \varepsilon_{j+1}(t)$ and $\varepsilon_{j+1}(t)$ under $a_{j+1}(t)$, $o_{j+1}(a_{j+1}(t))$, $u_{j+1}(t)$ and $x_{j+1}(t)$ (Appendix 2).

When $x'(t)$ is m classes of Fourier series with terms of r_1, r_2, \dots, r_m ($n = r_1 + r_2 + \dots + r_m$), it expresses Eq. 2.

$$x'(t) = \sum u_j(\tau_j(t))v_j(\tau_j(t)) + \varepsilon_m(t) \quad (j = 1, \dots, m) \tag{2}$$

In Eq. 2, $t = t(\tau_j)$ is the inverse scale transform of the j th scale τ_j to scale t , $v_j(t)$ is a Fourier series of the j th class modulation, and $\varepsilon_m(t)$ is the m th residual signal.

Equation 2 shows that $x'(t)$ is reconstructed completely by periodic Fourier series $v_j(\tau_j)$ on scale τ_j ($j = 1, 2, \dots, m$).

2.3 Quantity of information on the modulating signal

The relation between the quantity of information (3) and the modulation (4) is derived from formula (2).

Since the probability of components on the j th class modulation is to be $p_j = r_j/n$, the probability to encounter $x'(t)$ with m modulations is $p(r_1, r_2, \dots, r_m) = p_1^{r_1} \cdot p_2^{r_2} \cdot \dots \cdot p_m^{r_m}$ which relates to the geometrical mean $\sqrt[m]{p(r_1, r_2, \dots, r_m)}$. Equation is as follows (Appendix 5).

$$\sqrt[m]{p(r_1, r_2, \dots, r_m)} = \prod p_j^{p_j} = 1/m \tag{3}$$

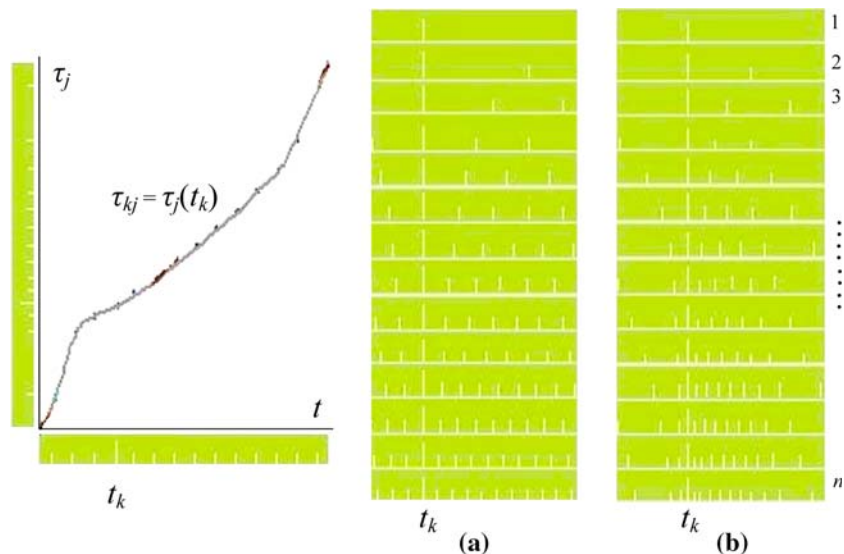
It clearly shows that the quantity of information on multi-modulating signal is m itself.

When $h_{ij}(\tau_j)$ is the i th instantaneous Fourier component of standard $x_j(\tau_j)$ in scale τ_j and $h_{ij}(t)$ is its modulating component in scale t , the morphological relation is given by the following Formula (4):

$$h_{ij}(t) = u_j(\tau_j)h_{ij}(\tau_j)\partial\tau_j/\partial t \tag{4}$$

In Eq. 4, $\partial\tau_j/\partial t$ is the Jacobian. Therefore, formula (4) shows that the modulation of every component at scale t is expressed by scale's modulations on its magnitude $u_j(\tau_j)$ and its time $\tau_j(t) = o_j(a(t))$. Especially, the adaptive scale $a_j(t)$ describes composite transform $a_j(t) = c_1(c_2(\dots c_n(t)))$ and that of exchanged orders $a_j(t) = c_n(\dots c_2(c_1(t)))$, so the s th scale $c_s(t)$ will express its specific modulation from the standard such as modulations of PQ, QRS, QT and RR.

Fig. 1 Synchronous frequency samplings on scale t and τ



3 Materials and methods

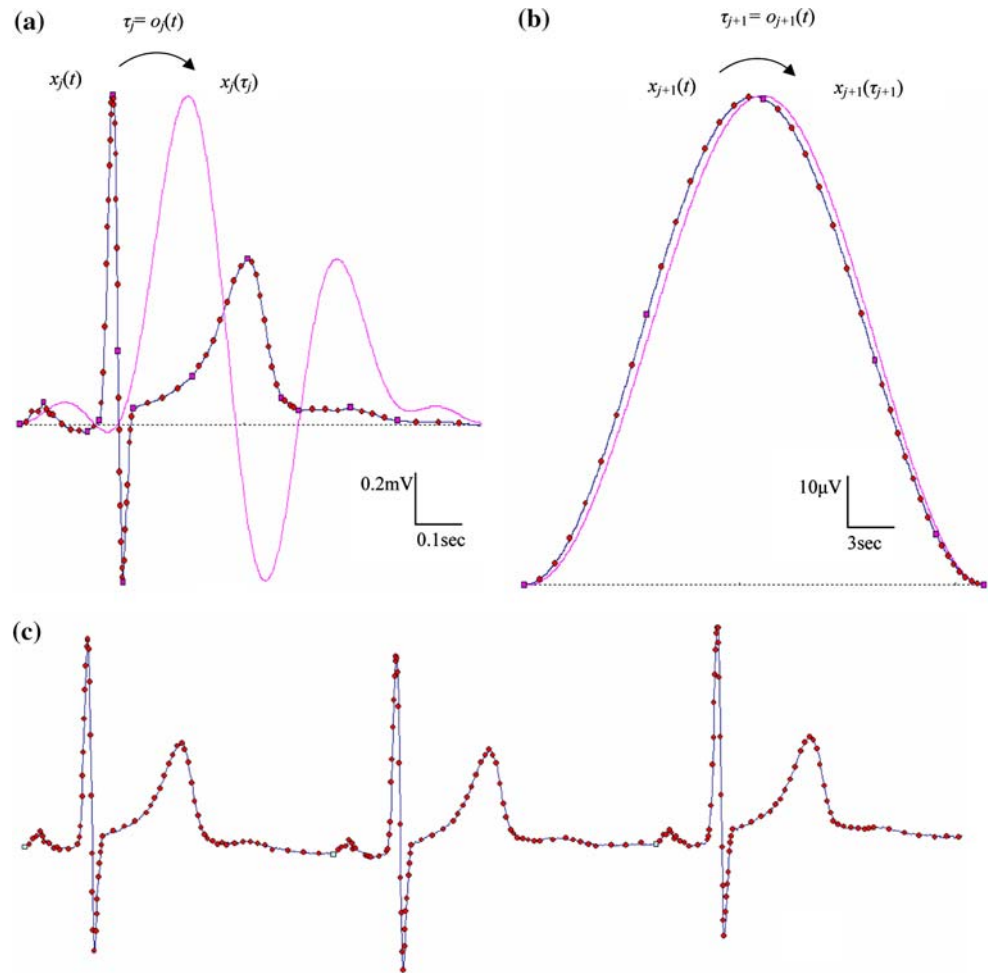
The analysis of scale classification of the three modulating signals on V4 lead was carried out by using the statistical Fourier analysis with maximum terms of 256 with A/D converted 16-bit samples of the sampling frequency 20 kHz over successive 50 s. To gather independent samples with 256 terms, samples needed to be accumulated by the high speed 20 kHz of A/D converter with a large dynamic range of 16 bit to avoid the deform of a tiny respiratory signal (Fig. 1).

The ECG standard $x_j(t)$ was selected from the source $x'(t)$ without breathing. On the other hand, $x'(t)$ with natural breathing and the respiratory flow $\dot{V}(t)$ are recorded simultaneously by using Number 3 Fleisch pneumotachograph. Both the ECG and the respiratory scale transforms were the same as in Fig. 2a, b of Appendix 2. The ECG $x'(t) - \varepsilon_j(t)$ and its baseline wander $\varepsilon_j(t)$ were analyzed when $u_j(\tau_j) = 1$ and the respiratory standard $x_{j+1}(t)$ was selected from stationary sequences of $\varepsilon_j(t)$ in 16 respiratory cycle. $(\varepsilon_j(\tau_{j+1}) - \varepsilon_{j+1}(\tau_{j+1}))/u_{j+1}(\tau_{j+1})$ and $\varepsilon_{j+1}(\tau_{j+1})/u_{j+1}(\tau_{j+1})$ were analyzed when $u_{j+1}(\tau_j)$ of the formula (5) in Appendix 4 and lung ventilation $V_{j+1}(t)$ and its residual $\varepsilon_{j+1}(t)$ were determined.

4 Results

Signals corresponding to every process on the scale classification of $m = 2$ under constant magnitude scale of $u_j = 1$ are illustrated in Fig. 3 with five cardiac periods. That is, intervals among specific points in source $x'(t)$ in Fig. 3a are transformed into the same intervals of the standard signal with the adaptive scale

Fig. 2 ECG and respiratory standard waves and specific points on scale t . **a** Cardiac standard, **b** respiratory standard, and **c** detecting dots on a part of the source



$a_j(t)$ to $x'(a_j(t))$ of Fig. 3b and with the inherent scale $\tau_j(t)$ to $x'(o_j(a_j(t)))$ of Fig. 3c. The transformed signal $x'(\tau_j)$ is decomposed into statistical Fourier components on a fundamental period of 53 cardiac intervals. The transformed signal $x'_j(\tau_j)$ in Fig. 3d is a reconstructed signal from the 53rd, 106th, 159th and 212th components, and its residual $\varepsilon_j(\tau_j)$ in Fig. 3e from all other components. The extracted ECG $x'_j(t)$ in Fig. 3f and the base line wander $\varepsilon_j(t)$ in Fig. 3g are the inverse scale-transformed signals from scale τ_j to scale t . When tiny-leak signals on the $\varepsilon_j(t)$ synchronized with intervals of QT are replaced with linear trends, the figure of pattern $\varepsilon_j(t)$ takes that of the lung ventilation in Fig. 3i. The smoothing signal, $V_j(t)$ in Fig. 3h is a filtered signal of $\varepsilon_j(t)$ with low-pass filter (LPF) under 30 terms (0.6 Hz). $V_j(t)$ is illustrated with enlarged four times. The signal $V(t)$ in Fig. 3i is the lung ventilation curve which is an integral signal of the direct measurement of the respiratory flow $\dot{V}(t)$.

Any differences of P ~ U waveforms between the source $x'(t)$ and reconstructed ECG $x'_j(t)$ bear getting information, and also the differences between ventila-

tion $V(t)$ and baseline wander $V_j(t)$ are small when frequencies lower than breathing are neglected. $x'_j(t)$ provides time-interval information on Minnesota codes better than that of $x'(t)$ clearly in visual.

Correlation coefficients ρ between the signal $V_j(t)$ and $V(t)$ are calculated by each ventilation with scatter plots in Fig. 4a, and are classified by P ~ U waves of 53 cardiac signals in Table 1a and by respiratory phases of 16 ventilations in Table 1b. Total ρ takes even values ($\rho \approx 0.82$) and has less interference between P ~ U waves in cardiac cycles. However, it takes low values ($\rho \leq 0.5$) at the transient regions of I, III between the inspiration and the expiration corresponding to loops in Fig. 4a. The inspiration and the expiration except transient regions are statistically dominant at higher value ($\rho \approx 0.95$). It is clear that the scale classification extracts the electrical activity of the heart with constant magnitude scale of $u_j(\tau_j) = 1$ in hole window of T_j . The extracted baseline wander is constructed with lung ventilation and other biological information.

The magnitude scale transform is applied to the scale classification of the base line wander $V_j(t)$ with a

Fig. 3 Classified process of modulating signal of the algorithm

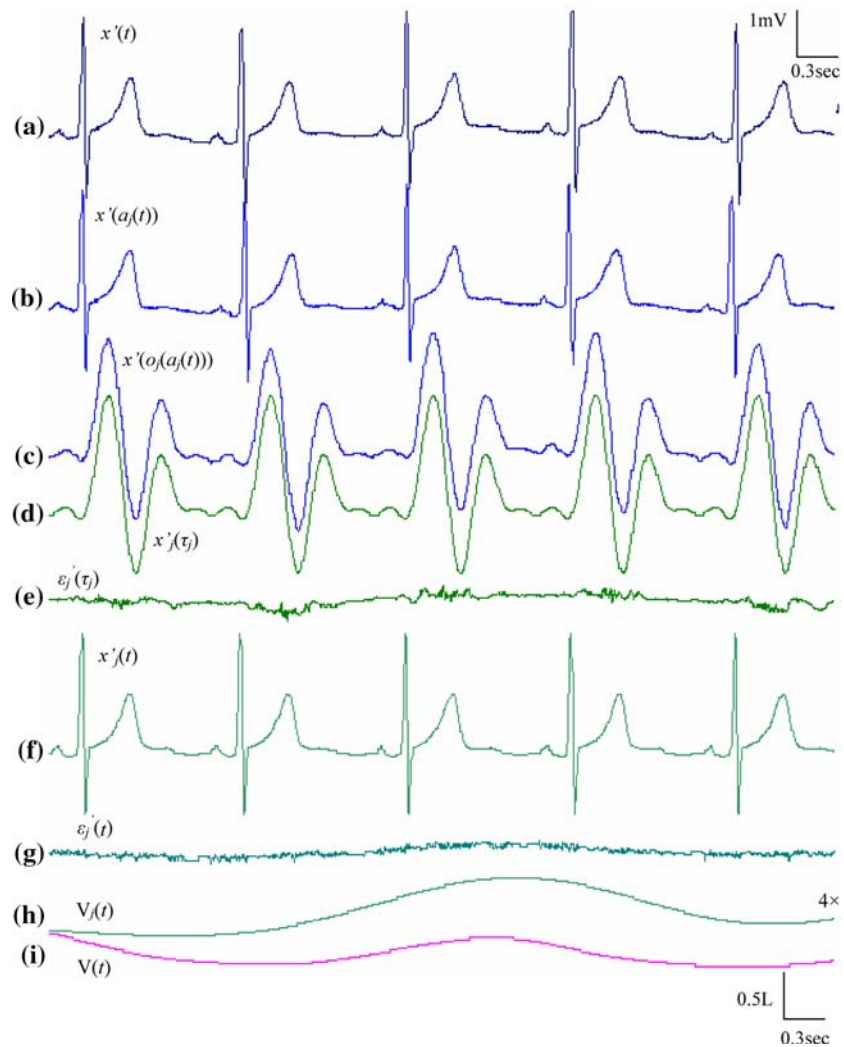
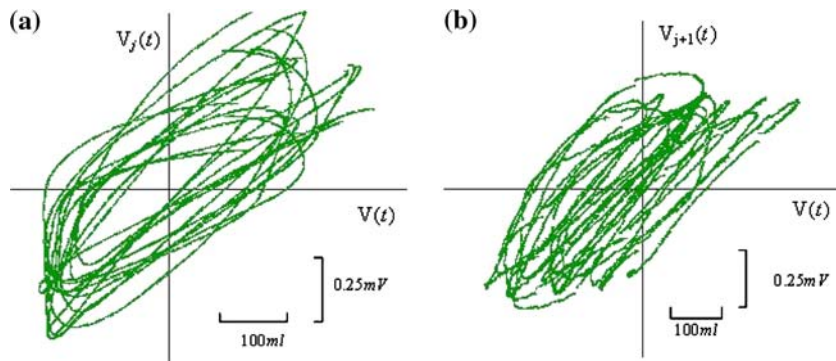


Fig. 4 X–Y loop among modulating $V_j(t)$, $V_{j+1}(t)$ and $V(t)$



respiratory standard in Fig. 2b. Scale classification of $m = 4$ is illustrated in Fig. 5, the source $x'(t)$ in Fig. 5a, the extracted ECG $x_j(t)$ in Fig. 5b, the extracted respiratory signal $V_{j+1}(t)$ in Fig. 5c, the residual $\epsilon_{j+1}(t)$ in Fig. 5d, the leak ECG $\epsilon(t)$ over 0.6 Hz of $\epsilon_j(t)$ in Fig. 5e and the ECG of $x_j(t) + \epsilon(t)$ in Fig. 5f. The signal $V(t)$ in Fig. 5g is 16 measured respiratory ventilations.

The magnitude scale $\Delta u_{j+1}(t)$ is illustrated in Fig. 5h. $V_{j+1}(t)$ and $\epsilon_{j+1}(t)$ are displayed in the reverse because the extracted respiratory waveform is a visually reversed $V(t)$. $\epsilon_{j+1}(t)$ is similar to base-line trends of $V(t)$ which relate to the thoracic gas volume (TGV).

The scatter plots of 15 lung ventilations between $V_{j+1}(t)$ and $V(t)$ take the similar pattern as illustrated in

Table 1 Correlation coefficients ρ_i among $V_j(t)$, $V_{j+1}(t)$ and $V(t)$ on cardiac and respiratory stages

Cardiac stages	PQ	QRS	ST-T	Base line	Total	
(a) Correlation coefficients ρ_i between $V_j(t)$ and $V(t)$ on ECG stages						
ρ_i	0.8328	0.8243	0.8149	0.8151	0.8244	
Number of samples	13,566	12,690	33,018	43,553	102,827	
Respiratory stages	I	II	III	VI	V	Total
(b) Correlation coefficients ρ_i between $V_j(t)$ and $V(t)$ on respiratory stages						
ρ_i	0.3990	0.9788	0.5772	0.9970	0.9546	0.7273
Number of samples	25,747	25,210	25,928	14,574	11,558	103,017
(c) Correlation coefficients ρ_i between $V_{j+1}(t)$ and $V(t)$ on respiratory stages						
ρ_i	0.9923	0.6589	0.9943	0.8032	0.7305	0.91648
Number of samples	25,286	22,816	21,163	15,155	18,077	102,497

I inspirations before midpoints, *II* inspirations after midpoints, *III* expirations before midpoints, *VI* expirations after midpoints, *V* expirations after inflection points

Fig. 5 Scale classification of the four modulating signals in an electro-cardiac signal

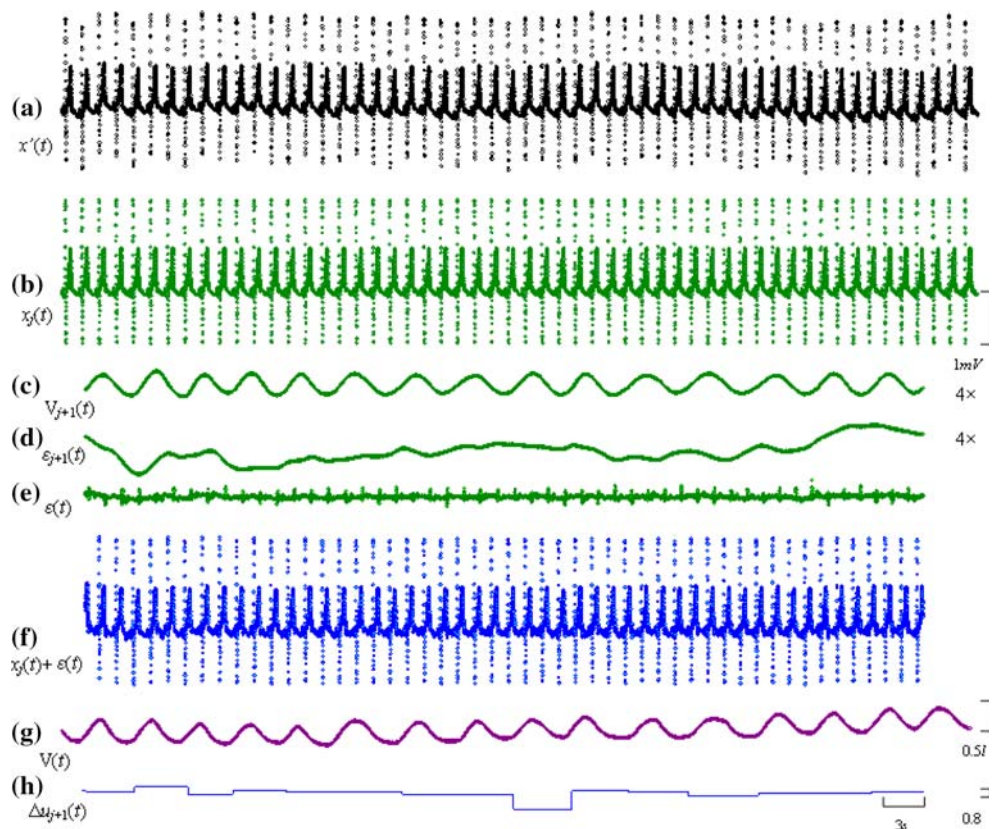


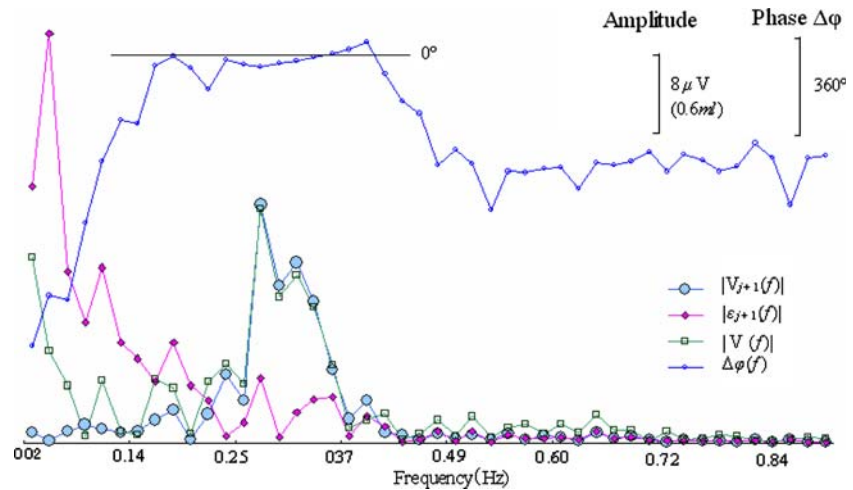
Fig. 4b. Correlation coefficients ρ in Table 1c takes higher value than the others though transient regions change I, III into II, IV.

The frequency distributions of the amplitude components on $|V(f)|$, $|V_{j+1}(f)|$, $|\epsilon_{j+1}(f)|$ and that of phase shifts $\Delta\phi(f)$ between $V(f)$ and $V_{j+1}(f)$ are illustrated in Fig. 6. On frequency ranges from 0.2 to 0.5 Hz, amplitude distribution in $|V(f)|$ and $|V_{j+1}(f)|$ takes similar pattern and those of phase shifts are invariant

near 0° . Amplitude distribution in $|\epsilon_{j+1}(f)|$ has not only the lower frequency components under 0.06 Hz corresponding to the trends of TGV but also the leak components of the lung ventilation from 0.25 to 0.5 Hz. The components of $|\epsilon_{j+1}(f)|$, moreover, has the harmonics among 0.04, 0.1, 0.18 and 0.28 Hz though those harmonics disappear in $|V(f)|$ and $|V_{j+1}(f)|$ at least.

As a result, the time and magnitude scale is effective to classify modulating signals from the higher

Fig. 6 Frequency distributions of the measured $V(f)$ and the extracted $V_{j+1}(f)$, $\varepsilon_{j+1}(f)$



frequency $V_{j+1}(t)$ to the lower frequency $\varepsilon_{j+1}(t)$ because a residual $\varepsilon_{j+1}(t)$ has less interference of the higher $V_{j+1}(t)$ and successive sequential $\varepsilon_{j+1}(t)$ will supply the signal for detecting other lower modulation.

5 Discussion and conclusion

A discrepancy between time domain and frequency domain analysis happens commonly by using biased samplings for discrete signal processing in various way such as Hilbert transform [2] and wavelet analysis [9] relating with an unique sampling frequency. Moreover, the mathematical concepts of one-to-one correspondence, n unknown amplitudes and phases are derived from $2n$ samples, hinders to orientate the statistics in the signal processing.

The SFT, however, has no contradiction between Fourier analysis and the statistics which resolute components with the same signal-to-noise ratio (SNR) for any signal decreasing amplitude components with order $1/f$ such as ECG signal, which is unsatisfied in the autoregressive spectral estimation [11]. The SFT component, moreover, is given independently by a time trace of instantaneous components at each instant as shown in Appendix 1. That is, its amplitude and phase are not calculated by the inner products between the source and trigonometric bases but by the extracted traces of instantaneous sin-wave themselves [12]. The concepts of signal arrangements at series $\{t_k\}$ is another thing on the SFT because the replacement between t_i and t_j needs the replacements of all the synchronous frequency samplings between t_i and t_j to hold the same instantaneous components. The summation of replacements, then, is about $(n - 1)!(n - 2)! \cdots 1$. In other words, the SFT supplies a signal ensemble arranging samples of the series $\{t_k\}$. In this study, dis-

crete samples from A/D converter of 20 kHz satisfy the independency as shown in Appendix 6. As a result, the random noise including a signal is not analyzed as a white noise but as inherent components characterizing arrangements. Therefore, the reconstructed waveform of a baseline wander holds a prototype even if it includes many random noises.

The necessary condition of the spectral analysis known as Nyquist’s condition [14, 15] becomes precise on the SFT as shown in Appendix 3, which confirms its condition statistically. The potentiality to scatter the aliased components impartially discriminates the waveforms under the unsatisfied Nyquist’s condition. In this algorithm, while the inherent scale transform converges to the standard $x_j(\tau_j)$ with four skipping components, others $\varepsilon_j(\tau_j)$, either continuous or random, disperse components into 256 terms.

Since the scale-transformed samples on the SFT are gathered from a whole window at any instant, those samples reflect its scale transform better than any local one. The random and leak signal on the extracted ECG $x'_j(t)$ are reduced into the tiny level under a stable behavior of the random noise. The 256 terms is enough to analyze scale-transformed signal in a window under 53 cardiac cycles because transformed components with the ranges, $8 \leq k_j \leq 53$, takes almost the same figures. The high-resolution analysis, therefore, is accomplished by the relative high SNR from those instantaneous components.

The problem on the scale transforms is to be adequate scales, not to be expressing the true biological modulation. $a_{j+1}(t)$ is, however, selected a traditional scale for non-stationary sequences in each respiratory cycle which carry information stored in Fourier phases as already indicated in [6] and $a_j(t)$ is selected from cardiac cycles without breathing in addition to the non-stationary sequences. As a result, $o_j(a_j)$ and $o_{j+1}(a_{j+1})$

make those mutual relationship of phases more simple relations. Taking own standard characterized with number of poles is also reasonable to determine the waveform reconstruction and to discriminate the own HRV. Moreover, the lost waveform is scattered into the other components without losing the morphologic information such as the leak ECG waveform.

It is well known the volume conductivity of the thorax takes inequality and changes the electro cardiac potentials on the thorax's surface with respiratory movements [1, 5].

In the scale classification of $m = 4$, it indicates that a constant magnitude scale $u_{kj} = 1$ produces unchangeable heart-activity $x'_j(t)$ so that the signal $\varepsilon_j(t)$ appears in respiratory modulations from their changes of the conductivity. It suggests that the standard without breathing follows the formula (2) of the signal classification and the source follows the formula (4) which expresses the modulating standard with Jacobian, $\partial\tau_j/\partial t$. A composite scale $c_s(t)$, therefore, is expected to express the inherent modulation from the standard such as PQ, QT and RR. That is, the spectral analysis of $c_s(t)$ on RR provides the distribution of the HRV without a tachogram which is expressed in other ways by others [2, 9, 11]. The extracted ECG has the potentiality to detect R–R modulations of the heart itself.

The algorithm extracts the lung ventilation curves $\{V_{j+1}(t)\}_k$ from V4 lead as changing the thoracic gas volume [1, 4]. The difference of amplitude distribution between $V(t)$ and $V_{j+1}(t)$ from spectral analysis takes the same pattern around 0.3 Hz [8]. The phase difference between $V(t)$ and $V_{j+1}(t)$, however, occurs in the transient regions between the inspiration and the expiration which deduce from inadequate respiratory standards, false respiratory peaks and the gas compression in the thorax [7]. It is effective to determine specific components, such as harmonics of $\varepsilon_{j+1}(t)$ around (0.04, 0.1, 0.18 and 0.28 Hz) in cardiovascular responses [2, 17].

Information of the partial ventilation is also expected from other 12 leads [8]. The respiratory Jacobian $\partial\tau_{j+1}/\partial t$, like that of ECG, has the potentiality to detect respiratory modulation itself as a part of the HRV and $u_{j+1}(t)$ is expected to detect the amplitude modulation such as sighs distribution. The residual $\varepsilon_{j+1}(t)$ supplies information of waveforms not only of the thoracic gas volume but also of the blood flow relating to cardiovascular responses [2, 11, 17].

It can be concluded that this algorithm reveals the simultaneous measurements of ECG and respiration. The comparison with the frequency distribution and

the waveform evaluates new modulations clearly, such as u_{kj+1} for detecting changes of the tidal volume (TV) and for finding TGV and cardiovascular responses. This contribution of the algorithm indicates diagnostic potentials on the sleep apnoea from ECG recordings [16].

6 Appendix 1

Instantaneous Fourier components on the SFT at any instant t_k are derived from samples of the synchronous frequency samplings from the source, having maximum components number of n , as illustrated in Fig. 1a. Those samples at any instant t_k are independent of each other when the t_k is given by $t_k = kT_j/s$ ($k = 0, \dots, s - 1$) in a window T_j for a prime number s of $s > 2n$. Although the sampling timing from the transformed source at scale τ_j is the same pattern in Fig. 1a, timing at scale t becomes as illustrated in Fig. 1b. Any source sample, therefore, is expressed by own components on each scale independently.

7 Appendix 2

The ECG standard $x_j(t)$ and its scale-transformed signal $x_j(\tau_j)$ are illustrated in Fig. 2a and those of the respiratory $x_{j+1}(t)$ and $x_{j+1}(\tau_{j+1})$ are illustrated in Fig. 2b. $x_j(\tau_j)$ is four terms of Fourier series passing eight poles of $x_j(t)$. $x_{j+1}(\tau_{j+1})$ is a sin wave passing two poles of $x_{j+1}(t)$. Inherent scale $\tau_j = o_j(t)$ and $\tau_{j+1} = o_{j+1}(t)$ are defined as transforms from $x_j(t)$, $x_{j+1}(t)$ into $x_j(\tau_j)$, $x_{j+1}(\tau_j)$.

The detected specific points for the source and the ECG standard are illustrated in Fig. 2a, c as a dot curve. These dots indicate start, pole and end of the P ~ U waves in each cardiac cycle. Specific points of lung ventilation are the poles of inspiration, expiration, midpoints and expiratory inflection point in each respiratory period. The midpoints represent respiratory maximum flows and expiratory inflection points distinguish expirations relating to pleural pressure. Points are determined to divide time interval of the specific-points into 6 ~ 16 to approximate true signals. Both the ECG and the respiratory standard are line curves of connected points. The adaptive scales $a_j(t)$, $a_{j+1}(t)$ are coordinated by divided points after adjusting detected specific points to those of the standard. The scale-transformed signal $x'(\tau_j)$ and that of the residual $\varepsilon_j(\tau_j)$ at scales τ_j , τ_{j+1} are determined from formula (2),

whereas the scale transforms are $\tau_j(t) = o_j(a_j(t))$ and $\tau_{j+1}(t) = o_{j+1}(a_{j+1}(t))$.

8 Appendix 3

The inherent transform $o_j(t)$ satisfies the necessary condition of the SFT analysis because Fourier components appear in the skipped ik_j th ($i = 1, \dots, r_j$) terms under $n > r_j k_j$ when $x'(\tau_j)/u_j(\tau_j)$ is periodic waves of k_j standard. Contrarily, instantaneous components are scattering into broadband of 1st ~ n th components when the source $x'(t)$ unfits scales τ_j, u_j .

The necessary condition of the SFT analysis is that the mean $m_i(t_k) (i = n + 1, \dots, 2n)$ over the $(n + 1)$ th of synchronous frequency sampling takes the same value at any instant t_k as $m_{n+1}(t_k) = m_{n+2}(t_k) = \dots = m_{2n}(t_k)$ which has the same properties with Nyquist's condition. Each instantaneous component includes errors of aliased components under the unsatisfied condition, but the summation of all the instantaneous components comes back to a source sample completely at any instant and the SNR of each component is improved with an adequate n value because the aliased errors are scattered into n components monotonously than those of the unique-interval samplings.

9 Appendix 4

The $\Delta u_{kj} (= u_{kj} - 1)$ is given by the following formula (5) when $x'_k(\tau_j) = v_{kj}(\tau_j) + \varepsilon_{kj}(\tau_j)$ under $u_{kj}(\tau_j) = 1$.

$$\Delta u_{kj} = (\varepsilon_{kj}(\tau_{mx}) - \varepsilon_{kj}(\tau_{mi})) / (v_{kj}(\tau_{mx}) - v_{kj}(\tau_{mi})) \quad (5)$$

$v_{kj}(\tau_{mx})$ and $v_{kj}(\tau_{mi})$ are the maximum and minimum values of $v_{kj}(\tau_j)$ when $u_{kj}(\tau_{mx}) = u_{kj}(\tau_{mi})$. That is, $\varepsilon_{kj}(\tau_j)$ takes $\varepsilon_{kj}(\tau_j) - \Delta u_{kj} v_{kj}(\tau_j)$ when $v_{kj}(\tau_j)$ changes $(1 + \Delta u_{kj}) v_{kj}(\tau_j)$ in a window T_{kj} .

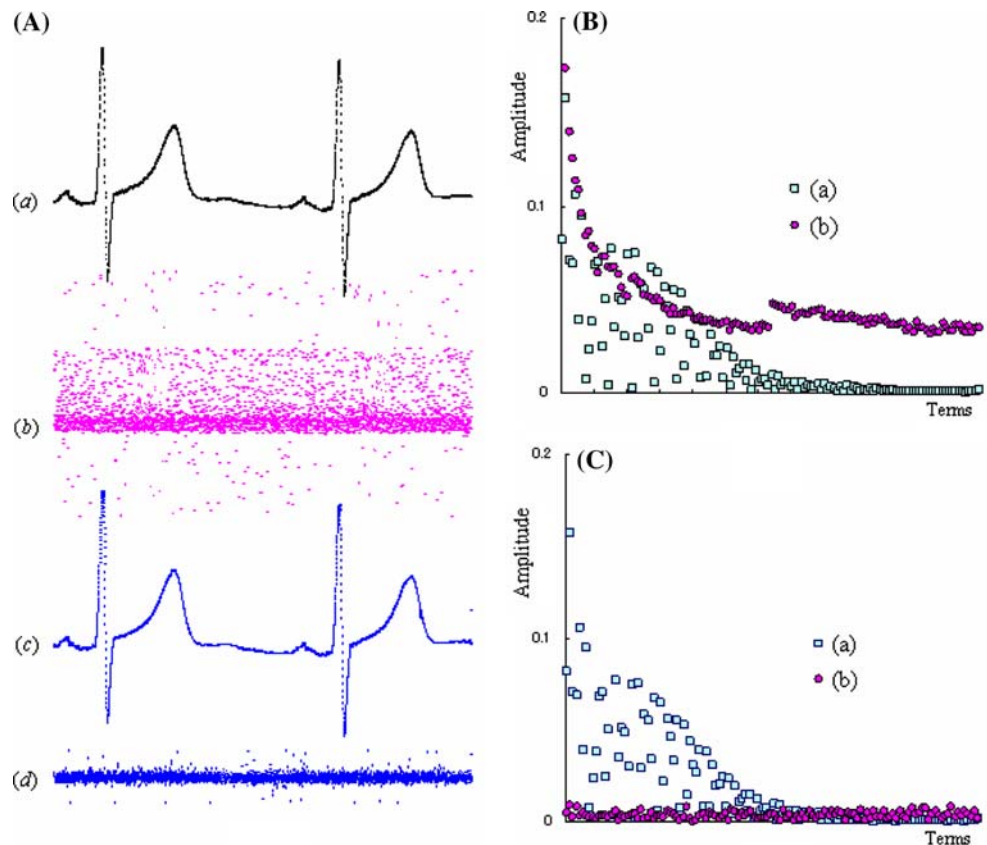
The constant component of each window T_{kj} is compensated under the continuous condition with an end expiratory point at each edge of window T_{kj} .

10 Appendix 5

The cases of modulations take $n! / r_1! r_2! \dots r_m!$ when the Fourier components of $n = r_1 + r_2 + \dots + r_m$ are distributed into m cases. The probability $p(r_1, r_2, \dots, r_m)$ to encounter $x'(t)$ with m modulations is given by Starling's formula as below.

$$p(r_1, r_2, \dots, r_m) = r_1! \cdot r_2! \cdot \dots \cdot r_m! / n! = p_1^{r_1} p_2^{r_2} \cdot \dots \cdot p_m^{r_m}$$

Fig. 7 The comparison between SFT and DFT on a randomized signal. **A** The reconstructed signals from SFT and DFT components on rearranged ECG, **B** amplitude distribution on the SFT, and **C** amplitude distribution on the DFT



$ms/n = 1$ is derived from the relation $\sum p_j = 1$ and $p_j = s/n$ when s is satisfied with the relation $\prod p_j^{p_j} = (s/n)^n$. That is, the geometrical mean s/n is satisfied with the relation of $s/n = \prod p_j^{p_j}$, so m is given as $m = \prod p_j^{-p_j}$.

11 Appendix 6

When the random signal in Fig. 7A-b is a permuted source in Fig. 7A-a, the amplitude distributions on the SFT are illustrated in Fig. 7B and on the direct Fourier transform (DFT) in Fig. 7C, respectively. While amplitude distributions of the source take the same forms between the SFT in Fig. 7B-a and the DFT in Fig. 7C-a, those of the random signal are different in Fig. 7B-b and C-b. That is, the distribution takes an exponential form on the SFT but takes an even form on the DFT which is well known as the form of a white noise.

As amplitude and phase components of the SFT reconstruct the random signal in Fig. 7A-b completely, the inverse rearrangement of the reconstructed random noise comes back to the source in Fig. 7A-c. Contrarily, the waveform from the DFT is still a random in Fig. 7A-d not the source. As a result, the distribution of samples is kept in SFT but lost in DFT. The SFT supplies true instantaneous Fourier components for any scale transform including arrangements. In other words, a signal with any random noise holds far more stable information in the SFT than in any other integral transform.

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